

NEAR-OPTIMAL FEEDBACK GUIDANCE
FOR AN ACCURATE
LUNAR LANDING

by

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A THESIS

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ABSTRACT

This research presents a novel guidance method for a lunar landing problem. The method facilitates efficiency and autonomy in a landing. The lunar landing problem is posed as a finite-time, fixed-terminal, optimal control problem. As a key finding of this work, the method of solution that is applied to construct the guidance mechanism employs a new extension of the State-Dependent Riccati Equation (SDRE) technique for constrained nonlinear dynamical systems in finite time. In general, the solution procedure yields a closed-loop control law for a dynamical system with point terminal constraints. Being a closed-loop solution, this SDRE technique calculates corrections for unpredicted external inputs, hardware errors, and other anomalies. In addition, this technique allows all calculations to be performed in real time, without requiring that gains be calculated a priori. This increases the flexibility to make changes to a landing in real time, if required.

The new SDRE-based feedback control technique is thoroughly investigated for accuracy, reliability, and computational efficiency. The pointwise linearization of the underlying SDRE methodology causes the new technique to be considered a suboptimal solution. To investigate the efficiency of the solution method, various numerical experiments are performed, and the results are presented. In addition, to validate the methodology, the new technique is compared with two other methods of solution: the Approximating Sequence of Riccati Equations (ASRE) technique and an indirect variational method, which provides the benchmark optimal open-loop solution.

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CONTENTS

ABSTRACT	ii
ACKNOWLEDGMENTS	iii
LIST OF TABLES	vii
LIST OF FIGURES	viii
CHAPTER 1. INTRODUCTION	1
CHAPTER 2. LITERATURE SURVEY.....	4
2.1 Apollo Guidance.....	4
2.2 Optimal Control Techniques	9
2.2.1 Techniques for Linear Systems.....	9
2.2.1.1 LQR Technique	9
2.2.1.2 Fixed-Final-State LQ Control	10
2.2.1.3 Indirect Variational Method	12
2.2.2 Techniques for Nonlinear Systems	14
2.2.2.1 SDRE Technique.....	14
2.2.2.2 Fixed-Final-State SDRE Technique.....	17
2.2.2.3 ASRE Technique.....	19

CHAPTER 3. DESCRIPTION OF THE LUNAR LANDING PROBLEM	22
CHAPTER 4. A NEW GUIDANCE METHOD FOR A LUNAR LANDING.....	24
4.1 Solution Description.....	24
4.2 Phase 1	25
4.2.1 Nondimensionalization of the Problem.....	26
4.2.2 Fixed-Final-State SDRE Solution for Phase 1	29
4.2.3 ASRE Solution for Phase 1	32
4.2.4 Indirect Variational Solution for Phase 1	33
4.3 Phase 2.....	35
CHAPTER 5. NUMERICAL RESULTS	37
5.1 Phase 1 Results.....	37
5.1.1 Experiment 1	37
5.1.2 Experiment 2	39
5.1.3 Experiment 3	41
5.1.4 Experiment 4.....	43
5.1.5 Comparison of the Techniques.....	45
5.2 Phase 2 Results.....	56
CHAPTER 6. CONCLUSIONS	58
REFERENCES	60

APPENDIX A. DESCRIPTION OF CONTROL SYSTEMS AND OPTIMAL CONTROL.....	63
APPENDIX B. DERIVATION OF THE DYNAMICAL SYSTEM	66

LIST OF TABLES

Table 1. Initial and final conditions for Phase 1	26
Table 2. Constants of the problem	27
Table 3. Variables of the problem.....	27
Table 4. Nondimensional variables of the problem	28
Table 5. Results of Experiment 1	38
Table 6. Results for the first part of Experiment 2.....	40
Table 7. Results for the second part of Experiment 2	41
Table 8. Results for the first part of Experiment 3.....	42
Table 9. Results for the second part of Experiment 3	43
Table 10. Matrices investigated in Experiment 4.....	44
Table 11. Results of Experiment 4.....	45
Table 12. Results of the three solutions	46
Table 13. Data for Phase 2	57
Table 14. Results of Phase 2	57

LIST OF FIGURES

Fig. 1. Apollo lunar landing diagram showing the various elements of a landing	5
Fig. 2. Apollo lunar landing phases	6
Fig. 3. Trajectory of a simulated Apollo landing	8
Fig. 4. Process for fixed-final-state LQ control	12
Fig. 5. Process of the SDRE technique	16
Fig. 6. Process of the fixed-final-state SDRE technique.....	18
Fig. 7. Diagram illustrating the simulation time steps	19
Fig. 8. Process of the ASRE technique	21
Fig. 9. Schematic of the lunar landing problem	22
Fig. 10. The new landing sequence	25
Fig. 11. Phase 1 landing trajectories of the various solutions	47
Fig. 12. Phase 1 landing trajectories of the various solutions, with thrust vectors	48
Fig. 13. Terminus of Phase 1 trajectories.....	49
Fig. 14. Terminus of Phase 1 trajectories, with thrust vectors	50
Fig. 15. Plot of r for the three solutions	50
Fig. 16. Plot of u for the three solutions.....	51
Fig. 17. Plot of v for the three solutions	51
Fig. 18. Plot of θ for the three solutions.....	52
Fig. 19. Plot of λ_r for the three solutions	53

Fig. 20. Plot of λ_u for the three solutions.....	53
Fig. 21. Plot of λ_v for the three solutions.....	54
Fig. 22. Plot of λ_θ for the three solutions.....	54
Fig. 23. Plot of U for the three solutions.....	55
Fig. 24. Plot of the angle of U , ϕ , for the three solutions.....	55
Fig. 25. Plot of cost-to-go, J , for the three solutions.....	56
Fig. 26. Diagram of a system	63
Fig. 27. Diagram of an open-loop control system.....	64
Fig. 28. Diagram of a closed-loop control system	64

CHAPTER 1

INTRODUCTION

Recently, there has been increased interest in returning to the Moon. However, it has been nearly forty years since humans have gone there. The last manned mission was in 1972, during the Apollo program. Since that time, a limited amount of research has been conducted to improve the landing methodology. A few research papers written on this matter are included in the References [1-6].

The Apollo missions had many limitations and had to overcome many challenges while using primitive computer technology. Landing sites were hazardous because they contained rocks and craters. Astronauts had to be able to adjust the landing point to avoid these hazards. This meant that the trajectory profile and the orientation of the lander had to allow the astronauts to observe the landing site. Moreover, due to the limitations of the computers, control systems were simplified as much as possible. These restrictions added inefficiency to the landings, thus resulting in wasted effort and fuel.

Apollo was successful in landing men on the Moon and returning them safely to the Earth. However, with the greater speed and power of present-day computer technology, it is highly desired to find an improved control method for landing on the Moon. Lunar stations of the near future will require a precise and efficient landing system that does not necessitate extensive training. The method investigated in this research could be the basis of such a desired system,

wherein a craft is allowed to perform all calculations in real time and is driven to a soft landing at a desired location while consuming a minimal amount of fuel. To achieve the autonomous goals of the new guidance method, lunar stations of the future would need to be clear of hazards and have very precise navigational aids. This would eliminate the need for astronauts to make course corrections and adjustments.

This research presents a novel guidance method for a lunar landing problem that is formulated as an optimal control problem. To obtain the feedback control law, the method of solution is presented using a new closed-loop feedback control technique for nonlinear systems with constraints. By performing various numerical experiments, the new guidance scheme is tested for its accuracy, efficiency, and robustness, and it is investigated with respect to computational burden. Through numerical simulations, the results for final location and velocity are compared with the desired values to determine the trajectory error and the precision of the control law. Validity of the solution is verified by comparing the results of this new technique with the results of two other optimal control solutions.

The research for this thesis is presented in the next five chapters. Chapter 2 is a literature survey, which covers the background material. In the first section of the chapter, the Apollo lunar landings are described. The second section discusses various common techniques for solving optimal control problems for linear and nonlinear systems. Finally, a new terminally constrained optimal control technique for nonlinear systems is presented.

Chapter 3 describes the modeling of the lunar landing problem. The chapter elaborates on all of the variables used in this research and includes a schematic of the problem. It also presents the equations of motion for the problem.

Chapter 4 describes a new guidance method. This chapter presents the complete sequence involved in the solution and describes the two phases of the method. Then it presents the new control technique along with other techniques applied in this new landing methodology.

Chapter 5 presents the results of simulations of the new guidance method. A thorough numerical analysis is demonstrated for various experiments with respect to different possible cases with the SDRE method. In addition, results comparing the SDRE method with two other methods are presented. Conclusions and future scope of this work are included in Chapter 6.

CHAPTER 2

LITERATURE SURVEY

This chapter covers background material for the problem considered in this research. The first section describes the Apollo lunar landing method, which includes the sequence of phases and the guidance solution used during a landing. The second section presents an overview of various optimal control techniques applied to solve linear and nonlinear systems. In this thesis, these techniques are respectively referred to as linear and nonlinear optimal control techniques, with respect to the dynamical system.

2.1 Apollo Guidance

The Apollo lunar landings were accomplished from 1969 to 1972, when computer technology was very primitive. The limitations of the computers made it impossible to utilize the advanced guidance techniques used today. Fig. 1 shows the overall landing procedure used by Apollo [7-10]. The lunar module (LM) separates from the command and service module (CSM) while in the parking orbit of approximately 60-nm in altitude. At a predetermined orbit position, the LM performs a Hohmann-type transfer maneuver [11]. The resulting elliptical orbit efficiently places the LM close to the Moon's surface in preparation for landing. At perilune, the LM fires its thrusters and begins the landing sequence.

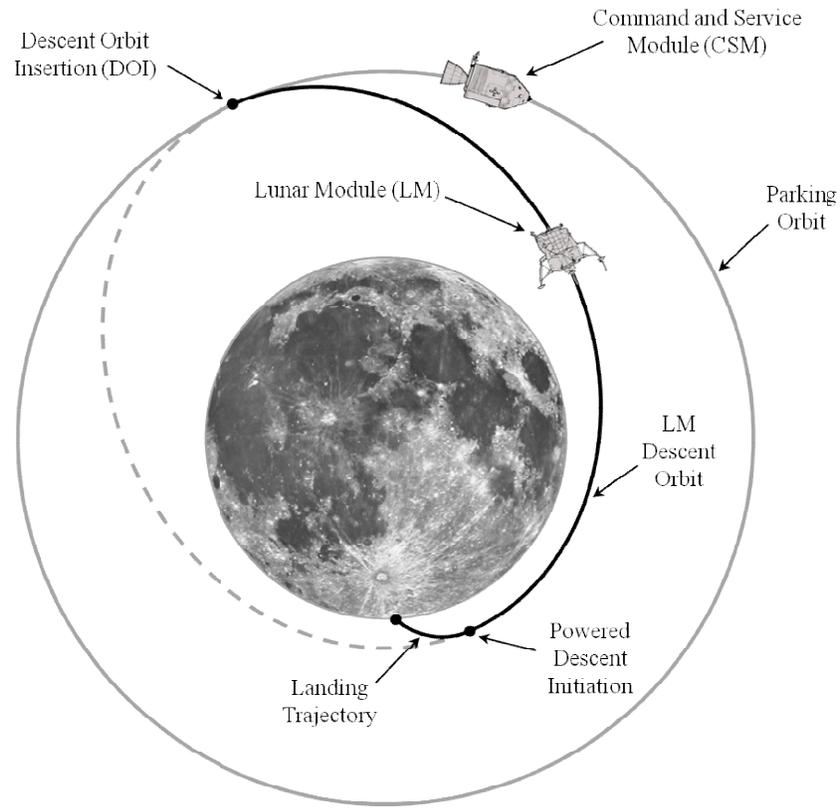


Fig. 1. Apollo lunar landing diagram showing the various elements of a landing

The landing sequence for Apollo consisted of three phases. As shown in Fig. 2, these are the braking phase (P63), the approach phase (P64), and the terminal descent phase (P66).

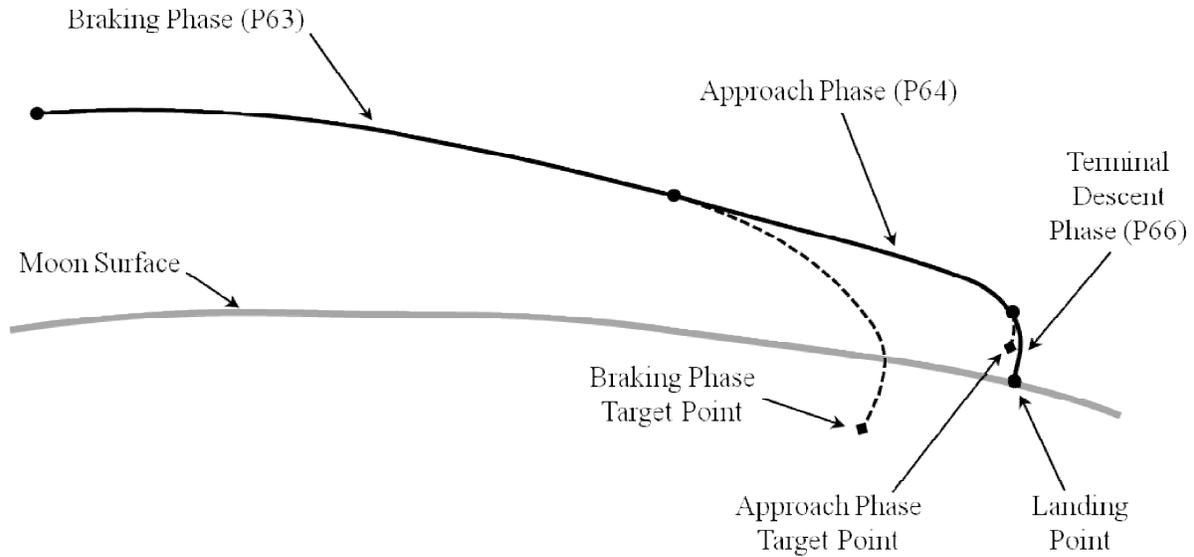


Fig. 2. Apollo lunar landing phases

P63 slows the LM from orbital speed. It typically begins at a 492-km slant range from the landing site and transfers the LM to the required initial conditions for P64. P64 begins immediately at the terminus of P63. Its objective is to deliver the LM to a point almost directly above the landing site. In addition, it provides continuous visibility of the lunar surface and of the landing location. This was a requirement in case the astronauts had to redesignate the landing point to avoid hazards. P66 typically begins automatically at a 30-m altitude. It provides velocity control but no position control. Forward and lateral velocities are nulled to produce a vertical approach to the landing site. The descent rate is controlled to a reference value that can be adjusted by the astronauts.

The trajectories for P63 and P64 were calculated prior to the Moon landings, because large ground-based computers were required to make the calculations. These trajectories were derived from a Taylor series expansion of the position function and are represented by a quartic polynomial [7] as

$$RRG = RTG + (VTG)T_C + (ATG)\frac{T_C^2}{2} + (JTG)\frac{T_C^3}{6} + (STG)\frac{T_C^4}{24} \quad (2.1)$$

where RRG is the position vector on the reference trajectory at current negative time T_C . RTG , VTG , ATG , JTG , and STG are the target position, velocity, acceleration, jerk, and snap vectors in guidance coordinates. P63 and P64 used separate sets of target values (calculated from a ground-based targeting program) to define their trajectories.

A quadratic guidance equation [7] was derived from Eq. (2.1). It is given as

$$ACG = \left[36\left(\frac{T_P}{T_C}\right)^2 - 24\left(\frac{T_P}{T_C}\right) \right] \frac{(RTG - RG)}{T_C^2} + \left[24\left(\frac{T_P}{T_C}\right)^2 - 18\left(\frac{T_P}{T_C}\right) \right] \frac{VTG}{T_C} + \left[12\left(\frac{T_P}{T_C}\right)^2 - 6\left(\frac{T_P}{T_C}\right) \right] \frac{VG}{T_C} + \left[6\left(\frac{T_P}{T_C}\right)^2 - 6\left(\frac{T_P}{T_C}\right) + 1 \right] ATG \quad (2.2)$$

where ACG , VG , and RG are the commanded acceleration, current velocity, and current position.

T_P is the predicted target-referenced time [7] defined as

$$T_P = T_C + Leadtime \quad (2.3)$$

where $Leadtime$ is the transport delay due to computation and command execution.

The current negative time T_C , or time-to-go, was calculated to satisfy the downrange Z-component of jerk. With a desired value for downrange jerk, JTG_z , the following equation [7] was numerically solved for T_C :

$$(JTG_z)T_C^3 + 6(ATG_z)T_C^2 + [18(VTG_z) + 6(VG_z)]T_C + 24[(RTG_z - RG_z)] = 0 \quad (2.4)$$

The calculated value for T_C was then used in Eq. (2.2) to determine the commanded acceleration required. Using Eq. (2.2) and target data for the Apollo 14 landing [10], a sample Apollo landing is simulated for this research. The resulting trajectory is shown in Fig. 3 and is shown in

the results of Chapter 5 for comparison with the other guidance techniques. The coordinate system shown in Fig. 3 is located at the center of the Moon.

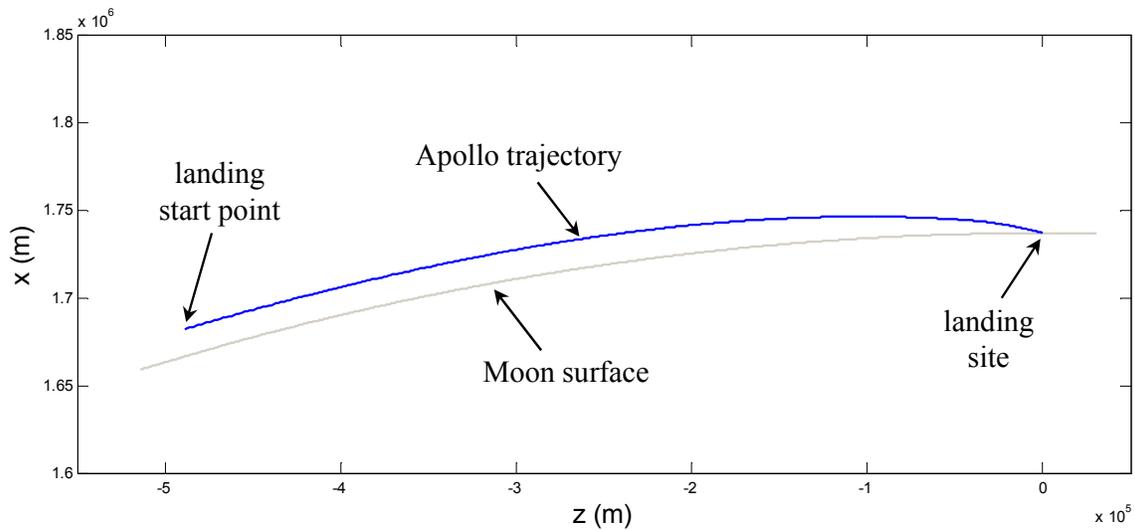


Fig. 3. Trajectory of a simulated Apollo landing

The Apollo program was successful in landing men on the Moon using the limited computers of the day. With today's advanced technology, a much better method can be employed. This would make it possible to land more precisely while being safer and more efficient. The method investigated in this research looks promising as being one that could be used to accomplish this goal.

This research is based on the use of optimal control theory to formulate a solution for the lunar landing problem. Appendix A presents background material on control systems and provides more details on optimal control. The appendix also gives the equation of a system in standard matrix form and shows diagrams of different types of systems. Further sections of this chapter cover various techniques for using optimal control theory for synthesizing the feedback control schemes. Some of these techniques form the basis for the research of this thesis.

2.2 Optimal Control Techniques

There are many techniques for solving optimal control problems (OCP), depending on whether the system is linear or nonlinear. Linear systems are easier to solve, and their solutions are well developed. Nonlinear systems are more difficult, and their solution techniques are not as mature. Some solutions for nonlinear systems are just now becoming realized, as computers are becoming powerful enough to be used in these systems effectively. The following sections describe some of the techniques used for linear and nonlinear dynamical systems.

2.2.1 Techniques for Linear Systems

To solve optimal control problems for linear systems, some of the popular techniques are the linear quadratic regulator (LQR) [12-17], the fixed-final-state linear quadratic (LQ) control method [13, 14], indirect variational methods [13, 14], and dynamic programming [12-14]. Since they are applied later in this thesis, LQR, fixed-final-state LQ control, and an indirect variational method are described in the following subsections.

2.2.1.1 LQR Technique

The linear quadratic regulator (LQR) technique gets its name from the fact that it operates on a linear system and a quadratic cost functional. The linear system has the state-space form [17] given as

$$\dot{\mathbf{x}} = \mathbf{Ax} + \mathbf{Bu} \quad (2.5)$$

With an LQR approach to obtain a closed-loop system, the control input $\mathbf{u} \in \mathbb{R}^m$, for full state feedback [17], is defined as

$$\mathbf{u} = -\mathbf{Kx} \quad (2.6)$$

where \mathbf{K} is the matrix of gains, and $\mathbf{x} \in \mathbb{R}^n$ is the state vector. The cost functional [17] is expressed as

$$J = \frac{1}{2} \int_0^{\infty} (\mathbf{x}^T \mathbf{Q} \mathbf{x} + \mathbf{u}^T \mathbf{R} \mathbf{u}) dt \quad (2.7)$$

The solution to an LQR problem is found by first solving the algebraic Riccati equation (ARE) [17] for the \mathbf{P} matrix prior to actually using the control system. The ARE for a linear system is given as

$$0 = \mathbf{A}^T \mathbf{P} + \mathbf{P} \mathbf{A} + \mathbf{Q} - \mathbf{P} \mathbf{B} \mathbf{R}^{-1} \mathbf{B}^T \mathbf{P} \quad (2.8)$$

The \mathbf{P} matrix, the solution of Eq. (2.8), is then used to calculate gain \mathbf{K} with

$$\mathbf{K} = \mathbf{R}^{-1} \mathbf{B}^T \mathbf{P} \quad (2.9)$$

Then, this gain matrix \mathbf{K} is used in Eq. (2.6) to compute control input \mathbf{u} at every sample time.

2.2.1.2 Fixed-Final-State LQ Control

Considering a soft constraint at the fixed-final-state with a linear system, the quadratic cost functional, or performance index [14], can be given as

$$J = \frac{1}{2} \mathbf{x}(T)^T \mathbf{S}(T) \mathbf{x}(T) + \frac{1}{2} \int_{t_0}^T (\mathbf{x}^T \mathbf{Q} \mathbf{x} + \mathbf{u}^T \mathbf{R} \mathbf{u}) dt \quad (2.10)$$

subject to

$$\mathbf{C} \mathbf{x}(T) = \mathbf{r}(T) \quad (2.11)$$

where \mathbf{C} is a constant coefficient matrix, and $\mathbf{x}(T)$ is the vector of final states. The states of a system are driven to a set of desired final values $\mathbf{r}(T)$ over a fixed amount of time, while minimizing the cost J .

For a full-state-feedback system, the following Riccati-type matrix differential equations [14] are needed to calculate the required input $\mathbf{u}(t)$ for a fixed-final-state LQ controller:

$$-\dot{\mathbf{S}} = \mathbf{A}^T \mathbf{S} + \mathbf{S} \mathbf{A} - \mathbf{S} \mathbf{B} \mathbf{R}^{-1} \mathbf{B}^T \mathbf{S} + \mathbf{Q}, \text{ given } \mathbf{S}(T) \quad (2.12)$$

$$\mathbf{K} = \mathbf{R}^{-1} \mathbf{B}^T \mathbf{S} \quad (2.13)$$

$$-\dot{\mathbf{V}} = (\mathbf{A} - \mathbf{B} \mathbf{K})^T \mathbf{V}, \text{ where } \mathbf{V}(T) = \mathbf{C}^T \quad (2.14)$$

$$\dot{\mathbf{P}} = \mathbf{V}^T \mathbf{B} \mathbf{R}^{-1} \mathbf{B}^T \mathbf{V}, \text{ where } \mathbf{P}(T) = 0 \quad (2.15)$$

$$\mathbf{u} = -(\mathbf{K} - \mathbf{R}^{-1} \mathbf{B}^T \mathbf{V} \mathbf{P}^{-1} \mathbf{V}^T) \mathbf{x} - \mathbf{R}^{-1} \mathbf{B}^T \mathbf{V} \mathbf{P}^{-1} \mathbf{r}(T) \quad (2.16)$$

In these equations, the values \mathbf{S} , \mathbf{K} , \mathbf{V} , \mathbf{P} , \mathbf{u} , and \mathbf{x} are functions of time t . The numerical process to compute the gains $\mathbf{K}(t)$, $\mathbf{V}(t)$, and $\mathbf{P}(t)$ for fixed-final-state LQ control is shown in Fig. 4. Given \mathbf{A} , \mathbf{B} , and final time T , the following steps are required in the precalculation before using the control system. With the prescribed values for $\mathbf{S}(T)$, \mathbf{Q} , \mathbf{R} , and Δt that will lead to desired results, the backward integration from T to t_0 is used to solve for $\mathbf{S}(t)$, using Eq. (2.12), then $\mathbf{S}(t)$ is used to find $\mathbf{K}(t)$ in Eq. (2.13). Now, using backward integration from T to t_0 , $\mathbf{V}(t)$ and $\mathbf{P}(t)$ are solved using Eqs. (2.14) - (2.15). Since the control system is running in real time, the stored values of $\mathbf{K}(t)$, $\mathbf{V}(t)$, and $\mathbf{P}(t)$, along with the current state vector \mathbf{x} and the desired final state vector $\mathbf{r}(T)$, are used to find the required control input vector \mathbf{u} at every sample time, using Eq. (2.16).

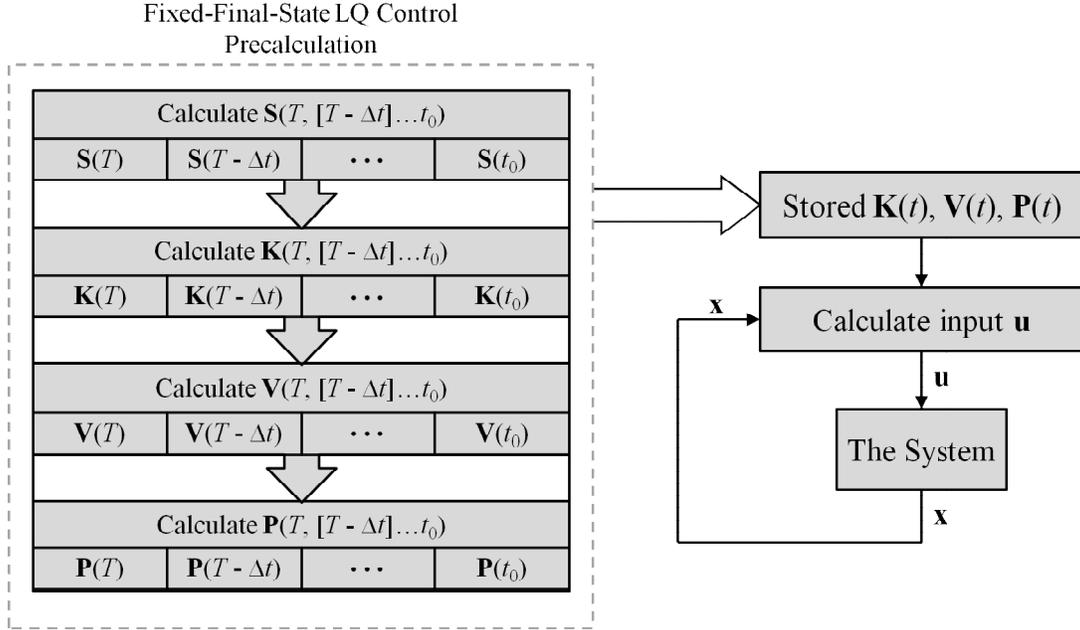


Fig. 4. Process for fixed-final-state LQ control

2.2.1.3 Indirect Variational Method

In an indirect variational method, the calculus of variations approach is used to derive Euler-Lagrange equations and the optimality conditions for solving optimal control problems. The results of this method are considered optimal, subject to satisfying the second order necessary conditions for optimality.

For a fixed-final-state problem with a system described as

$$\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}, \mathbf{u}, t) \quad (2.17)$$

a cost functional for a variational method [14] has the form

$$J = \varphi(\mathbf{x}(T), T) + \int_{t_0}^T L(\mathbf{x}, \mathbf{u}, t) dt \quad (2.18)$$

where φ represents the soft constraint that includes a weighting function on the final state at the final time, and L is a function of state and input at intermediate times along the trajectory to be considered for optimization. The terminal constraint [14] is defined as

$$\boldsymbol{\psi}(\mathbf{x}(T), T) = 0 \quad (2.19)$$

with $\boldsymbol{\psi}$ being zero-valued expressions for the final state. A Hamiltonian equation [14] can be constructed as

$$H(\mathbf{x}, \mathbf{u}, t) = L(\mathbf{x}, \mathbf{u}, t) + \boldsymbol{\lambda}^T \mathbf{f}(\mathbf{x}, \mathbf{u}, t) \quad (2.20)$$

where $\boldsymbol{\lambda}$ is the set of Lagrange multipliers, also known as costates. The first order necessary conditions for optimality are used to derive the differential equations for costate, stationarity condition, and transversality condition [14]. Respectively, these are given as

$$-\dot{\boldsymbol{\lambda}} = \frac{\partial H}{\partial \mathbf{x}} \quad (2.21)$$

$$0 = \frac{\partial H}{\partial \mathbf{u}} \quad (2.22)$$

$$\left(\boldsymbol{\varphi}_x + \boldsymbol{\Psi}_x^T \mathbf{v} - \boldsymbol{\lambda} \right)^T \Big|_T dx(T) + \left(\boldsymbol{\varphi}_t + \boldsymbol{\Psi}_t^T \mathbf{v} + H \right) \Big|_T dT = 0 \quad (2.23)$$

These three equations, with the given initial conditions on the states, create a two-point boundary value problem (TPBVP). The solution of the TPBVP can be attempted by starting with initial guesses for $\boldsymbol{\lambda}(0)$ and \mathbf{v} and then using a shooting method [18, 19] to solve for the values of $\boldsymbol{\lambda}(t)$ and \mathbf{v} , which satisfy the boundary conditions of the problem. From $\boldsymbol{\lambda}(t)$ and expressions derived from Eq. (2.22), the required input $\mathbf{u}(t)$ can be calculated.

There are some negative aspects to this technique. It is an open-loop control law, and all the input values $\mathbf{u}(t)$ have to be calculated prior to using the control system. Also, the shooting

method for solving the TPBVP is problematic. If the initial guesses are poor, then accurate results may not be produced, or convergence of the solution may not occur at all for nonlinear systems.

2.2.2 Techniques for Nonlinear Systems

There are various techniques used to solve optimal control problems for nonlinear systems. Some of these are the State-Dependent Riccati Equation (SDRE) method [20-34], the Approximating Sequence of Riccati Equations (ASRE) technique [25, 35, 36], indirect variational methods, and nonlinear programming (NLP) techniques [13, 37].

This research focuses on application of SDRE for finite time optimal control problems and compares the solution process with ASRE and an indirect variational method. Basic descriptions of SDRE and ASRE are given in the following subsections.

2.2.2.1 SDRE Technique

The State-Dependent Riccati Equation (SDRE) technique is a relatively new solution method for solving nonlinear optimal control problems. The most common use of the technique is as a regulator based on the LQR methodology described in Section 2.2.1.1, using a quadratic cost functional such as Eq. (2.7).

The goal of the method is to solve nonlinear optimal control problems in feedback form by approximating the system as linear at each sample time. The linear system at each of these points in time is based on the current states at that time. At each sample time, the problem is solved using common LQR techniques. The assumption of pointwise linearization creates a suboptimal solution, but it facilitates in solving a difficult nonlinear system for a feedback

control. Moreover, with enough sample points along the trajectory, the suboptimal solution can be made to be very close to optimal.

The heart of the SDRE strategy is to factor the nonlinear model into a linear-like form [20] as

$$\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}, \mathbf{u}, t) = \mathbf{A}(\mathbf{x})\mathbf{x}(t) + \mathbf{B}(\mathbf{x})\mathbf{u}(t) \quad (2.24)$$

In this form, \mathbf{A} and \mathbf{B} are functions of \mathbf{x} , and their numerical values change throughout the trajectory of \mathbf{x} . There are many possibilities for the form of $\mathbf{A}(\mathbf{x})$ and $\mathbf{B}(\mathbf{x})$. For example, the system

$$\dot{\mathbf{x}} = \begin{bmatrix} \dot{c} \\ \dot{d} \end{bmatrix} = \begin{bmatrix} (c + d^2) \\ (d + c/d) \end{bmatrix}$$

can be factored into

$$\overbrace{\begin{bmatrix} 1 & d \\ (1/d) & 1 \end{bmatrix}}^{\mathbf{A}} \overbrace{\begin{bmatrix} c \\ d \end{bmatrix}}^{\mathbf{x}} \quad \text{or} \quad \overbrace{\begin{bmatrix} (1 + d^2/c) & 0 \\ (d/c) & (c/d^2) \end{bmatrix}}^{\mathbf{A}} \overbrace{\begin{bmatrix} c \\ d \end{bmatrix}}^{\mathbf{x}}$$

The forms for $\mathbf{A}(\mathbf{x})$ and $\mathbf{B}(\mathbf{x})$ have to be decided to give the desired results. If the best matrix forms are not known in advance, then a variety of these matrices can be tested in simulations in order to facilitate the decision. Moreover, before distinct $\mathbf{A}(\mathbf{x})$ and $\mathbf{B}(\mathbf{x})$ matrices are used in the SDRE formulation, they should be tested for controllability [12, 16, 17] over the entire state trajectory. $\mathbf{A}(\mathbf{x})$ and $\mathbf{B}(\mathbf{x})$ are controllable if

$$\begin{bmatrix} \mathbf{B} & \mathbf{A}\mathbf{B} & \mathbf{A}^2\mathbf{B} & \dots & \mathbf{A}^{n-1}\mathbf{B} \end{bmatrix} \quad (2.25)$$

has full rank of n for all applicable values of state \mathbf{x} . If $\mathbf{A}(\mathbf{x})$ and $\mathbf{B}(\mathbf{x})$ are not controllable over the entire state trajectory, then the form of the matrices should be changed. After the final

choices for these matrices have been decided, they can be used in the SDRE technique to calculate required control input $\mathbf{u}(t)$ in real time.

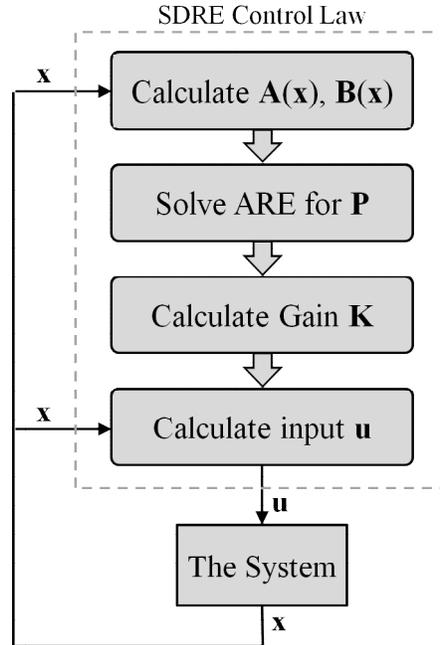


Fig. 5. Process of the SDRE technique

Fig. 5 shows a diagram of the SDRE technique. At each sample time, the following procedure is accomplished. First, the current state vector \mathbf{x} is used to calculate numerical values for $\mathbf{A}(\mathbf{x})$ and $\mathbf{B}(\mathbf{x})$. Then, using the LQR equations, \mathbf{P} and \mathbf{K} are calculated. Input \mathbf{u} is then calculated and applied to the system. This procedure is then repeated at the next sample time.

For the SDRE technique, the ARE is solved at every sample time for each new value of $\mathbf{A}(\mathbf{x})$ and $\mathbf{B}(\mathbf{x})$. This causes the nonlinear system to be approximated as a series of linear systems. Therefore, shorter time increments increase the accuracy of the control law, because this decreases the amount of time that each approximation is applied.

Because of its approximating nature, the SDRE technique is considered a suboptimal solution. However, with the proper choices for the $\mathbf{A}(\mathbf{x})$ and $\mathbf{B}(\mathbf{x})$ matrices, and with the proper amount of sample times, the SDRE technique can provide a very adequate solution.

For this research, a new form of the SDRE technique is formulated for the finite-time OCP posed with terminal constraints. The details of this new terminally constrained technique are given next.

2.2.2.2 Fixed-Final-State SDRE Technique

This new technique is derived from the aforementioned SDRE method. The normal SDRE method is most commonly used to control nonlinear systems over an infinite horizon, as described above. The new SDRE technique described in this section can be used to solve finite-horizon, terminally constrained, optimal control problems.

This reformulated SDRE technique combines the common SDRE method with the linear fixed-final-state LQ control solution of Section 2.2.1.2. This produces a fixed-final-state SDRE solution for nonlinear optimal control.

In general, the new technique operates the same as that of the more common SDRE regulator. At every sample time, it recalculates the $\mathbf{A}(\mathbf{x})$ and $\mathbf{B}(\mathbf{x})$ matrices and then calculates the input based on current states. The exception is that, instead of using the LQR strategy and solving the algebraic Riccati equation, the new technique uses the fixed-final-state LQ control strategy. Like the fixed-final-state LQ control technique, the values for $\mathbf{S}(T)$, \mathbf{Q} , and \mathbf{R} have to be chosen to give the best desired results. In addition, a value for the time increment Δt should be properly chosen. A value of Δt that is too long will reduce the accuracy of the control law, but one that is too short could over burden the system.

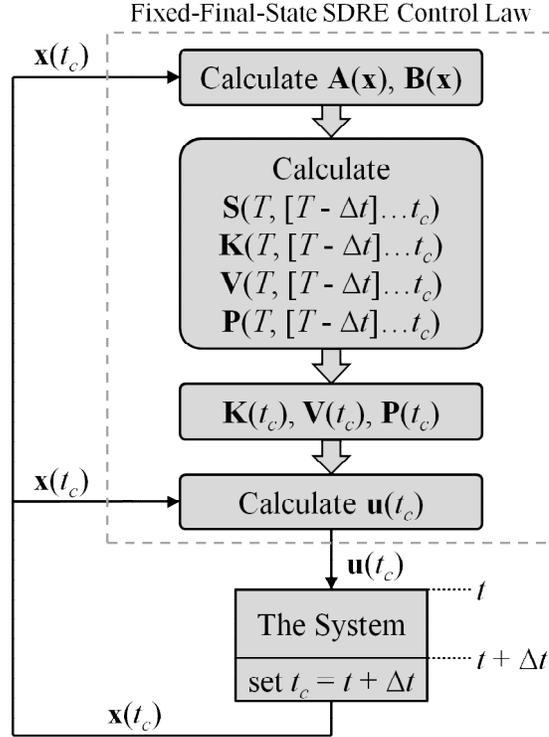


Fig. 6. Process of the fixed-final-state SDRE technique

Fig. 6 describes how this new control law works. As the figure shows, the following steps occur at every sample time while the control system is running. First, at the current time t_c , the matrices $\mathbf{A}(\mathbf{x})$ and $\mathbf{B}(\mathbf{x})$ are calculated from the current state $\mathbf{x}(t_c)$. Next, the values for $\mathbf{S}(t)$, $\mathbf{K}(t)$, $\mathbf{V}(t)$, and $\mathbf{P}(t)$ are calculated using Eqs. (2.12) - (2.15), with backward integration from final time T to current time t_c . The values $\mathbf{K}(t_c)$, $\mathbf{V}(t_c)$, and $\mathbf{P}(t_c)$ are used in Eq. (2.16), along with the current state vector $\mathbf{x}(t_c)$ and the desired final state vector $\mathbf{r}(T)$, to calculate the required input $\mathbf{u}(t_c)$ for the current sample time. The input $\mathbf{u}(t_c)$ is then applied to the system until the next sample time. At the next sample time, a new value for $\mathbf{x}(t_c)$ is measured, and the process is repeated. The iterations continue until the final time T has been reached.

This new SDRE formulation could be altered in many ways. One alteration would be to use a variable sampling rate. The sampling rate could start slow and then get faster over time. This

could reduce computational burden on the system. Another alteration could be to precalculate and record all of the $\mathbf{K}(t)$, $\mathbf{V}(t)$, and $\mathbf{P}(t)$ values in an SDRE simulation. These recorded gain values could then be used to make a much faster control system. In addition, an open-loop system could be created by recording all of the calculated input values in an SDRE simulation.

2.2.2.3 ASRE Technique

The Approximating Sequence of Riccati Equations (ASRE) technique is another solution for solving nonlinear optimal control problems iteratively. It can be used to solve terminally constrained optimal control problems over a fixed amount of time.

As with the SDRE technique, ASRE first factors the nonlinear system into the linear-like form of Eq. (2.24). However, ASRE is an iterative solution that changes the linear-like form into [25]

$$\dot{\mathbf{x}}^{[i]}(t) = \mathbf{A}(\mathbf{x}^{[i-1]}(t))\mathbf{x}^{[i]}(t) + \mathbf{B}(\mathbf{x}^{[i-1]}(t))\mathbf{u}^{[i]}(t) \quad (2.26)$$

where i represents the current iteration number. The technique is based on fixed-final-state LQ control. It has a cost functional like Eq. (2.10), and it utilizes the Eqs. (2.12) - (2.16).

Prior to using the control system, a precalculation has to be performed. This precalculation is an iterative approach that involves a simulation of the system, where the simulation time $t \in [t_0, T]$ is divided into a determined amount of time steps. Fig. 7 shows the relationships between t , $\mathbf{u}(t)$, $\mathbf{x}(t)$, and the time steps.

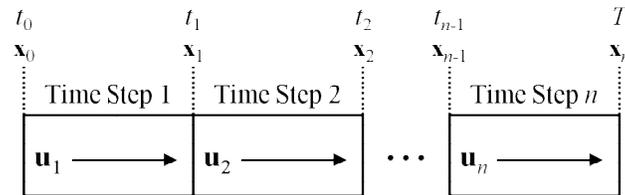


Fig. 7. Diagram illustrating the simulation time steps

Fig. 8 shows a diagram of how the ASRE technique works. The first part of the precalculation is to calculate $\mathbf{A}(\mathbf{x})$ and $\mathbf{B}(\mathbf{x})$ using the initial state vector \mathbf{x}_0 . The resulting values \mathbf{A}_0 and \mathbf{B}_0 are then used at every time step. The next part begins the iteration loop of the technique. The \mathbf{A} and \mathbf{B} values from every time step are used with Eqs. (2.12) - (2.15) and backward integration from $\mathbf{S}(T)$, $\mathbf{V}(T)$, and $\mathbf{P}(T)$ to calculate \mathbf{S} , \mathbf{K} , \mathbf{V} , and \mathbf{P} for every time step. Then, by using Eq. (2.16) with a simulation of the system, \mathbf{u} and \mathbf{x} are calculated for every time step. The values for \mathbf{x} are then used to calculate new values for \mathbf{A} and \mathbf{B} at every time step. The new values for \mathbf{A} and \mathbf{B} are now used in the next iteration of the loop. The loop is iterated a fixed number of times or until the final state error reaches a value below a set threshold. The calculated gain values $\mathbf{K}(t)$, $\mathbf{V}(t)$, and $\mathbf{P}(t)$ of the final iteration are recorded. These recorded gains can then be used in the closed-loop control system to calculate required input \mathbf{u} , as shown in Fig. 8. Or, by storing the input values $\mathbf{u}(t)$ of the final iteration of the precalculation, an open-loop control system can be created.

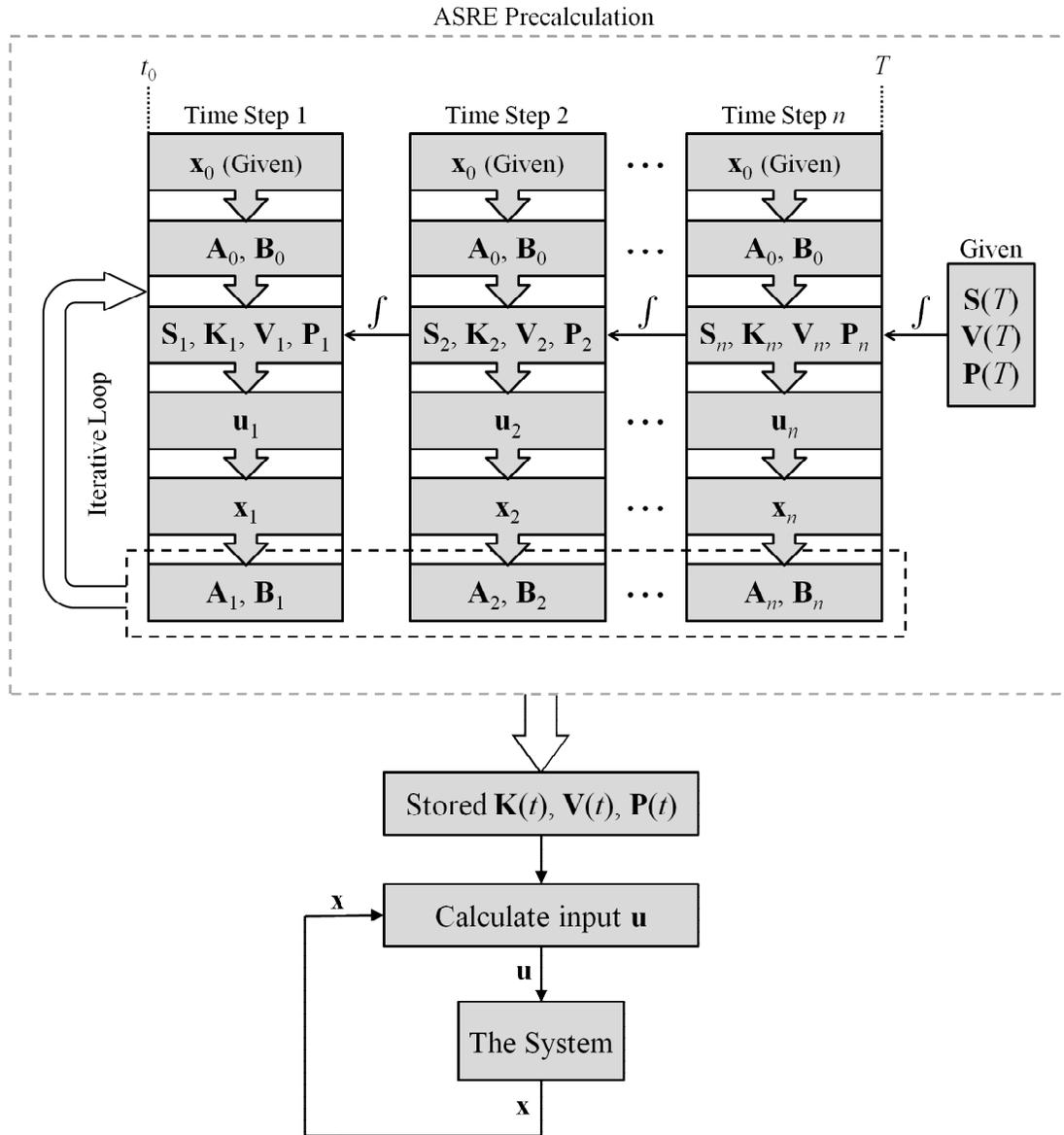


Fig. 8. Process of the ASRE technique

The ASRE technique, like SDRE, can be used to find a good suboptimal solution to a difficult nonlinear problem. It is noticed in this research that the ASRE technique is more reliable and is easier to implement than the indirect variational technique of Section 2.2.1.3. ASRE does not require very close guess values, and it converges reliably within only a few iterations.

CHAPTER 3

DESCRIPTION OF THE LUNAR LANDING PROBLEM

This chapter describes the lunar landing problem that is investigated in this research. Fig. 9 illustrates the variables used in the lunar landing problem. The lander is shown as a black dot in a clockwise orbit around the Moon, represented by the large circle. RF_1 and RF_2 are the two coordinate reference frames used in the landing solution. RF_1 is located at the center of the Moon, and RF_2 is located at the landing site. The radial distance from RF_1 to the lander is represented by r , and the positional angle is represented by θ . The radial and tangential velocities of the lander are depicted as u and v , respectively. U is the input acceleration due to applied thrust. U_r and U_t are the radial and tangential components of this input acceleration. The angle of U from the z-axis is represented by ϕ .

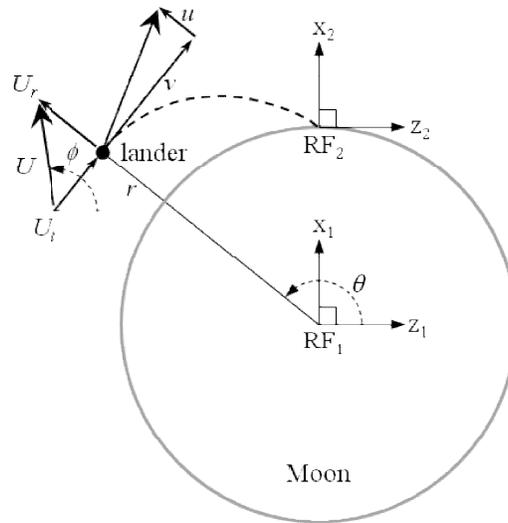


Fig. 9. Schematic of the lunar landing problem

The dynamical system equations [14] used for the lunar landing problem are

$$\dot{r} = u \tag{3.1}$$

$$\dot{u} = \frac{v^2}{r} - \frac{\mu}{r^2} + U_r \tag{3.2}$$

$$\dot{v} = -\frac{uv}{r} + U_t \tag{3.3}$$

$$\dot{\theta} = -\frac{v}{r} \tag{3.4}$$

In these equations, the variables \dot{r} and $\dot{\theta}$ are the radial and angular velocities, respectively; whereas, the variables \dot{u} and \dot{v} are the radial and tangential accelerations, respectively. The derivation of these equations is included in Appendix B.

CHAPTER 4

A NEW GUIDANCE METHOD FOR A LUNAR LANDING

This chapter describes the method of solution used to solve the lunar landing problem. It illustrates and defines the phases of the landing, and it describes optimal control techniques applied in the solution procedure of the guidance method.

4.1 Solution Description

The solution procedure is constructed in two phases, Phase 1 and Phase 2. The initial and final conditions for the two phases are chosen to match the Apollo missions closely. Phase 1, as the lander decelerates from orbit, utilizes a nonlinear optimal control technique. It minimizes fuel consumption as it works to stop the lander at a target point 30 meters above the landing location. This target point acts as a safety zone that allows for some small error in the nonlinear control. Phase 2 reformulates the dynamical system for the final portion of the landing into an approximate linear system and then uses linear optimal control to drive the vehicle down to a soft pinpoint landing.

Fig. 10 shows a schematic of the landing sequence. For solving the nonlinear Phase 1 portion of the problem, various nonlinear optimal control techniques are considered. For this research, it is decided to use a new form of the SDRE technique, as described in Section 2.2.2.2. The results of this new technique are compared to those obtained from the ASRE and variational methods described previously. All three of these solutions are simulated, and the resulting numerical data

is presented for comparison. The Apollo technique is also simulated to compare its trajectory to the trajectories of the other three solutions.

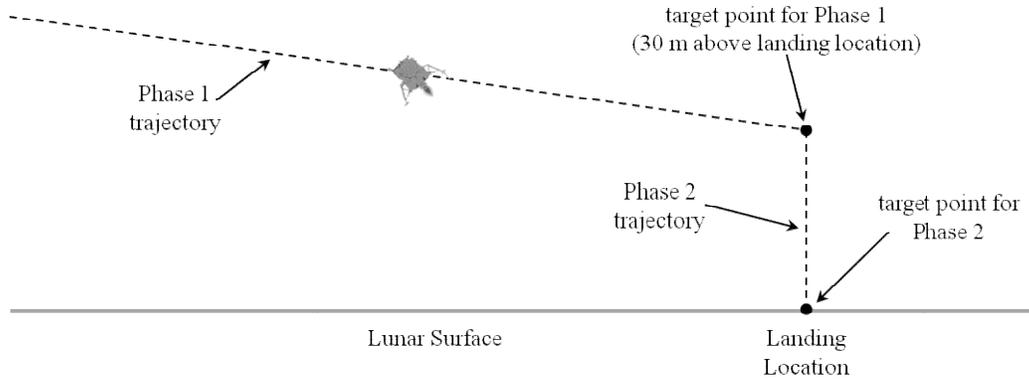


Fig. 10. The new landing sequence

4.2 Phase 1

Phase 1 covers the first portion of the landing, where orbital velocity and distance from the Moon cause significant nonlinearity in the system. It takes the spacecraft from orbit and brings it to rest 30 meters above the desired landing point on the Moon, as shown in Fig. 10. This phase uses the RF_1 reference frame shown in Fig. 9.

Table 1 shows the initial and final conditions used by Phase 1. The value for r_0 is determined for a 15-km initial altitude, the value r_f is determined for a 30-m final altitude, and θ_0 is calculated for a 492-km initial slant range to the landing site.

Table 1. Initial and final conditions for Phase 1

Parameter	Description	Value	Units
t_0	Initial time	0	(min)
t_f	Final time	11	(min)
r_0	Initial radial distance	1752100	(m)
r_f	Final radial distance	1737000	(m)
u_0	Initial radial velocity	0	(m/s)
u_f	Final radial velocity	0	(m/s)
v_0	Initial tangential velocity	1673	(m/s)
v_f	Final tangential velocity	0	(m/s)
θ_0	Initial position angle	1.85362	(radians)
θ_f	Final position angle	$\pi/2$	(radians)

For this research, a Moon radius r_m of 1,737 km, a Moon mass of 73.48×10^{21} kg, and a Moon gravitational parameter μ of $4903 \text{ km}^3/\text{s}^2$ are used [11]. In addition, for thrust calculations, a lander mass of 16,430 kilograms is estimated.

4.2.1 Nondimensionalization of the Problem

For Phase 1 of the lunar landing problem, it is beneficial to nondimensionalize the system. Nondimensionalization is the process of removing all units of measure from the equations by multiplying and dividing various constants. For problems that contain large numerical values, nondimensionalization can bring about dramatic improvement in calculation time. Sometimes, it can make a problem solvable, where before it was not. Nondimensionalization improves the results of this research. The process on how this is performed is given below.

First, two constants are chosen to be the basis for nondimensionalizing the problem. These are shown, along with their units, in Table 2. They are chosen because of their large values.

Table 2. Constants of the problem

Constant	Description	Units
μ	Gravitational Parameter	(m ³ /s ²)
r_m	Radius of the Moon	(m)

The variables of the problem are then identified, along with their units. These are shown in Table 3.

Table 3. Variables of the problem

Variable	Description	Units
r	Radial Distance	(m)
u	Radial Velocity	(m/s)
v	Tangential Velocity	(m/s)
θ	Positional Angle	(radians)
t	Time	(s)
U	Input Acceleration	(m/s ²)

Nondimensional representations of these variables are then formulated using the constants of Table 2. These new nondimensional forms are shown in Table 4.

Table 4. Nondimensional variables of the problem

Nondimensional Variable	Value
\bar{r}	$\frac{r}{r_m}$
\bar{u}, \bar{v}	$(u, v) \left(\frac{\mu}{r_m} \right)^{-1/2}$
$\bar{\theta}$	θ
\bar{t}	$t \left(\frac{\mu}{r_m^3} \right)^{1/2}$
\bar{U}	$U \left(\frac{\mu}{r_m^2} \right)^{-1}$

After rearranging the values of Table 4, differentiating, and using various substitutions, the following expressions for the derivatives of the states are formulated:

$$\dot{r} = r_m \left(\frac{d\bar{r}}{d\bar{t}} \right) \left(\frac{\mu}{r_m^3} \right)^{1/2}$$

$$\dot{u} = \left(\frac{\mu}{r_m} \right)^{1/2} \left(\frac{d\bar{u}}{d\bar{t}} \right) \left(\frac{\mu}{r_m^3} \right)^{1/2}$$

$$\dot{v} = \left(\frac{\mu}{r_m} \right)^{1/2} \left(\frac{d\bar{v}}{d\bar{t}} \right) \left(\frac{\mu}{r_m^3} \right)^{1/2}$$

$$\dot{\theta} = \left(\frac{d\theta}{d\bar{t}} \right) \left(\frac{\mu}{r_m^3} \right)^{1/2}$$

Then, by making further rearrangements and substitutions, the nondimensional forms of Eqs. (3.1) - (3.4) are found. These new nondimensional state equations are

$$\dot{\bar{r}} = \bar{u} \quad (4.1)$$

$$\dot{\bar{u}} = \frac{\bar{v}^2}{\bar{r}} - \frac{1}{\bar{r}^2} + \bar{U}_r \quad (4.2)$$

$$\dot{\bar{v}} = -\frac{\bar{u}\bar{v}}{\bar{r}} + \bar{U}_t \quad (4.3)$$

$$\dot{\bar{\theta}} = -\frac{\bar{v}}{\bar{r}} \quad (4.4)$$

These equations, along with the nondimensional values of Table 4, are used in the solution techniques for Phase 1. This allows the computations to run faster and with fewer problems.

4.2.2 Fixed-Final-State SDRE Solution for Phase 1

This section describes how the new fixed-final-state SDRE technique, described in Section 2.2.2.2, is used to solve Phase 1 of the lunar landing problem. First, the nonlinear state equations of the problem, given as Eqs. (3.1) - (3.4), are formulated into the linear-like form of Eq. (2.24). To do this, the elements \mathbf{x} , \mathbf{u} , $\mathbf{A}(\mathbf{x})$, and $\mathbf{B}(\mathbf{x})$ have to be defined. The first element \mathbf{x} , being the state vector for Phase 1, is defined as

$$\mathbf{x} \equiv \begin{bmatrix} r \\ u \\ v \\ \theta \end{bmatrix} \quad (4.5)$$

The input vector \mathbf{u} is defined to be

$$\mathbf{u} \equiv \begin{bmatrix} U_r \\ U_t \end{bmatrix} \quad (4.6)$$

For this problem, \mathbf{B} is a constant coefficient matrix given as

$$\mathbf{B} = \begin{bmatrix} 0 & 0 \\ 1 & 0 \\ 0 & 1 \\ 0 & 0 \end{bmatrix} \quad (4.7)$$

Defining $\mathbf{A}(\mathbf{x})$ is a little more difficult. Given that there are many possibilities for this matrix, various forms have to be evaluated in order to make the decision. The evaluated matrices are first tested for controllability. Then, several of the controllable forms are separately used in a simulation to see which gives the best results. Experiment 1, in Section 5.1.1, shows the results of simulations using various $\mathbf{A}(\mathbf{x})$ matrices. From the results of the evaluations, it is decided to use the form

$$\mathbf{A}(\mathbf{x}) = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -\frac{\mu}{r^3} & 0 & \frac{v}{r} & 0 \\ 0 & -\frac{v}{r} & 0 & 0 \\ 0 & 0 & -\frac{1}{r} & 0 \end{bmatrix} \quad (4.8)$$

Section 5.1.1 provides an explanation for this decision.

The nondimensional form of the $\mathbf{A}(\mathbf{x})$ matrix, corresponding to nondimensional state equations of Eqs. (4.1) - (4.4), is defined as

$$\mathbf{A}(\bar{\mathbf{x}}) = \begin{bmatrix} 0 & 1 & 0 & 0 \\ -1/\bar{r}^3 & 0 & \bar{v}/\bar{r} & 0 \\ 0 & -\bar{v}/\bar{r} & 0 & 0 \\ 0 & 0 & -1/\bar{r} & 0 \end{bmatrix} \quad (4.9)$$

Next, the values for the weighting matrices $\mathbf{S}(T)$, \mathbf{Q} , and \mathbf{R} are defined. For the Phase 1 solution, it is determined to use the values

$$\mathbf{S}(T) = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}, \quad \mathbf{Q} = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}, \quad \mathbf{R} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$$

The zero values for $\mathbf{S}(T)$ and \mathbf{Q} mean that the final states and the intermediate states along the trajectory are not minimized. For this particular problem, all states are being driven to fixed final values; therefore, there is no need to minimize them. The \mathbf{R} weighting matrix is used to minimize the input \mathbf{u} along the trajectory. It is decided that the unit matrix given to \mathbf{R} , as shown, will suffice for this research.

The desired final states of Phase 1 are shown in Table 1. The vector $\mathbf{r}(T)$, for these desired final states, includes all of the elements of the final state vector $\mathbf{x}(T)$. Therefore, $\mathbf{r}(T)$ is given as

$$\mathbf{r}(T) = \mathbf{x}(T) \equiv \begin{bmatrix} r_f \\ u_f \\ v_f \\ \theta_f \end{bmatrix} = \begin{bmatrix} 1737000 \text{ m} \\ 0 \text{ m/s} \\ 0 \text{ m/s} \\ \pi/2 \text{ radians} \end{bmatrix} \quad (4.10)$$

Therefore, the value of matrix \mathbf{C} , in Eq. (2.11), has to be the unit matrix given as

$$\mathbf{C} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix} \quad (4.11)$$

This \mathbf{C} matrix defines $\mathbf{V}(T)$ in Eq. (2.14).

The values given above, along with the initial state values shown in Table 1, are used with the new fixed-final-state SDRE technique of Section 2.2.2.2 to formulate a solution for Phase 1. During the process, the relationships of Table 4 are used to nondimensionalize the desired final state vector $\mathbf{r}(T)$, the current state vector $\mathbf{x}(t_c)$, the final time T , and the time increment Δt . A

fourth-order adaptive-step-size Runge-Kutta integrator is used to solve for $\mathbf{S}(t)$, $\mathbf{V}(t)$, and $\mathbf{P}(t)$ in Eqs. (2.12), (2.14), and (2.15). The resulting nondimensional input value \mathbf{u} , calculated from Eq. (2.16), is then dimensionalized and used in the system simulation to find the next current state vector $\mathbf{x}(t_c)$. In order to simulate the system, the same type of Runge-Kutta integrator is used to solve for state in Eqs. (3.1) - (3.4).

At the end of the simulation, the value J in Eq. (2.10) is calculated using all of the stored values for $\mathbf{x}(t)$ and $\mathbf{u}(t)$. The value of the integral within this equation is found by using a simple Euler method.

Costate λ , defined in Section 2.2.1.3, is calculated for this fixed-final-state SDRE solution in order to compare it to the costate of the variational solution in Section 4.2.4. To calculate the costate [13], a value for \mathbf{v} is found first using

$$\mathbf{v}(t) = \mathbf{P}^{-1}(t) \left[\mathbf{r}(T) - \mathbf{V}^T(t) \mathbf{x}(t) \right] \quad (4.12)$$

Then, costate is calculated from this value using the following equation:

$$\lambda(t) = \mathbf{S}(t) \mathbf{x}(t) + \mathbf{V}(t) \mathbf{v}(t) \quad (4.13)$$

4.2.3 ASRE Solution for Phase 1

The ASRE technique, described in Section 2.2.2.3, is also used to solve Phase 1 of the lunar landing. The purpose for the ASRE solution is to provide results that can be compared against the SDRE solution. This comparison is used to validate the SDRE solution and to provide a level of confidence that the solution is worth considering for future missions to the Moon.

The same values used in the SDRE solution are used with the ASRE technique to create the ASRE solution to the landing problem. As in the previous section, J and $\lambda(t)$ are calculated for

this solution in the same manner. All of the results of simulations using this solution are included in Chapter 5 for comparison with the SDRE results.

4.2.4 Indirect Variational Solution for Phase 1

An indirect variational technique is also used for Phase 1, and its results are considered optimal. As with ASRE, the variational solution provides results that are used to compare with those from SDRE. This adds yet another level of confidence to the SDRE solution.

This variational solution uses the technique described in Section 2.2.1.3. The cost functional for this solution is the same as Eq. (2.10). Using this cost functional form with the optimality equations of the technique, the optimal control law equations [14] for this solution are formulated to be

$$-\dot{\lambda} = \frac{\partial \mathbf{f}^T}{\partial \mathbf{x}} \lambda + \mathbf{Qx} \quad (4.14)$$

$$0 = \mathbf{Ru} + \frac{\partial \mathbf{f}^T}{\partial \mathbf{u}} \lambda \quad (4.15)$$

The state equations of the system, Eqs. (3.1) - (3.4), are used with Eq. (4.14) to create the following costate equations:

$$\dot{\lambda}_r = \left(\frac{v^2}{r^2} - \frac{2\mu}{r^3} \right) \lambda_u - \left(\frac{uv}{r^2} \right) \lambda_v - \left(\frac{v}{r^2} \right) \lambda_\theta \quad (4.16)$$

$$\dot{\lambda}_u = -\lambda_r + \left(\frac{v}{r} \right) \lambda_v \quad (4.17)$$

$$\dot{\lambda}_v = \left(-\frac{2v}{r} \right) \lambda_u + \left(\frac{u}{r} \right) \lambda_v + \left(\frac{1}{r} \right) \lambda_\theta \quad (4.18)$$

$$\dot{\lambda}_\theta = 0 \quad (4.19)$$

Equation (4.15) is then used to formulate equations for input \mathbf{u} . These are calculated to be

$$U_r = -\lambda_u \quad (4.20)$$

$$U_t = -\lambda_v \quad (4.21)$$

The vector of zero-valued expressions for the final state is defined as

$$\boldsymbol{\Psi} = \begin{bmatrix} r_f - 1737000 \\ u_f \\ v_f \\ \theta_f - \frac{\pi}{2} \end{bmatrix} \quad (4.22)$$

This is used in Eq. (2.23) to give the following expression:

$$\boldsymbol{\lambda} = \begin{bmatrix} \lambda_r(T) \\ \lambda_u(T) \\ \lambda_v(T) \\ \lambda_\theta(T) \end{bmatrix} = \begin{bmatrix} v_1 \\ v_2 \\ v_3 \\ v_4 \end{bmatrix} \quad (4.23)$$

It is extremely difficult to find proper guess values for $\boldsymbol{\lambda}(0)$ and \mathbf{v} . Without proper guess values, the solver will not produce good results. To simplify the process, the values for $\boldsymbol{\lambda}(0)$ and \mathbf{v} calculated from the SDRE solution are used for the guess values in the variational solution.

Using the initial values for the states and the guess values mentioned above, a numerical solver is used to solve the state equations, solve the costate equations, and find the values of $\boldsymbol{\lambda}(0)$ and \mathbf{v} that drive the final state conditions to the desired values. The state and costate equations are then integrated using the initial state values and the calculated initial costate values to find $\mathbf{x}(t)$ and $\boldsymbol{\lambda}(t)$. Then, the input $\mathbf{u}(t)$ is calculated from $\lambda_u(t)$ and $\lambda_v(t)$ using Eqs. (4.20) and (4.21).

As with the solutions of the previous two sections, the cost value J is calculated in the same manner for this solution. This J value is included in the results of Chapter 5.

4.3 Phase 2

Phase 2 drives the lander down vertically to the surface of the Moon, as shown in Fig. 10. It approximates the system as linear, assumes gravity constant, and uses linear fixed-final-state LQ control, as described in Section 2.2.1.2. This phase uses the RF₂ reference frame that is shown in Fig. 9.

The initial conditions for Phase 2 are the final conditions of Phase 1, transformed into RF₂ coordinates. The final conditions for Phase 2 are all zero relative to this reference frame.

The state equations of the approximate linear system are defined as

$$\dot{z} = v_z \quad (4.24)$$

$$\dot{x} = v_x \quad (4.25)$$

$$\dot{v}_z = U_z \quad (4.26)$$

$$\dot{v}_x = U_x^* \quad (4.27)$$

where positional coordinates are represented by x and z , and the respective velocities are represented by v_x and v_z . U_z is the z -component of the thrust input acceleration. U_x^* is the total input acceleration, including gravity, in the x -direction. U_x , defined as

$$U_x = U_x^* + g \quad (4.28)$$

is the x -component of the thrust input acceleration. For this equation, a calculated gravity value g of 1.625 m/s² is used.

The matrix form of the state equations can be represented as

$$\dot{\mathbf{x}} \equiv \begin{bmatrix} \dot{z} \\ \dot{x} \\ \dot{v}_z \\ \dot{v}_x \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}}_{\mathbf{A}} \underbrace{\begin{bmatrix} z \\ x \\ v_z \\ v_x \end{bmatrix}}_{\mathbf{x}} + \underbrace{\begin{bmatrix} 0 & 0 \\ 0 & 0 \\ 1 & 0 \\ 0 & 1 \end{bmatrix}}_{\mathbf{B}} \begin{bmatrix} U_z \\ U_x^* \end{bmatrix} \quad (4.29)$$

The \mathbf{C} constant coefficient matrix and the vector of desired final states $\mathbf{r}(T)$ for Phase 2 are defined to be

$$\mathbf{r}(T) \equiv \begin{bmatrix} z_f \\ x_f \\ v_{zf} \\ v_{xf} \end{bmatrix} = \begin{bmatrix} 0 \text{ m} \\ 0 \text{ m} \\ 0 \text{ m/s} \\ 0 \text{ m/s} \end{bmatrix}, \quad \mathbf{C} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \\ 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}$$

Also, the $\mathbf{S}(T)$, \mathbf{Q} , and \mathbf{R} matrices are given as

$$\mathbf{S}(T) = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}, \quad \mathbf{Q} = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}, \quad \mathbf{R} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$$

All of these values are used with the technique of Section 2.2.1.2 to find the necessary $\mathbf{u}(t)$ that drives the lander to a soft pinpoint landing at the surface of the Moon.

Phase 2 is used to complete the landing sequence for this research and to help validate the overall approach. Since the technique used for this phase is common and well known, and since this final portion of the landing is only a small part of the sequence, less emphasis is placed on Phase 2. However, the results for this phase are included in Chapter 5.

CHAPTER 5

NUMERICAL RESULTS

This chapter shows the results of simulations of the lunar landing method investigated in this research. Various numerical experiments are conducted on the new fixed-final-state SDRE solution for Phase 1. These experiments are described in this chapter, and their results are presented. In addition to these experiments, the results of the various solutions for Phase 1 are compared together to add validity to the fixed-final-state SDRE solution. The results of the Phase 2 solution are also shown in this chapter.

5.1 Phase 1 Results

The following subsections describe the numerical simulations that are performed for Phase 1. The results of each experiment are also given within the corresponding subsections.

5.1.1 Experiment 1

As discussed in Section 2.2.2.1, the $\mathbf{A}(\mathbf{x})$ matrix, for the factored system form given by Eq. (2.24), can be constructed in many ways. In this experiment, various forms of the $\mathbf{A}(\mathbf{x})$ matrix for the fixed-final-state SDRE solution, shown in Section 4.2.2, are used in separate simulations to compare their results. The purpose of this is to show how changing the form of the $\mathbf{A}(\mathbf{x})$ matrix affects the results of the solution. The results are studied to decide which $\mathbf{A}(\mathbf{x})$ matrix to use for the remainder of the research.

The following $\mathbf{A}(\mathbf{x})$ matrices are investigated in this experiment:

$$\mathbf{A}(\mathbf{x})_1 \equiv \begin{bmatrix} 0 & 1 & 0 & 0 \\ -\mu/r^3 & 0 & v/r & 0 \\ 0 & -v/r & 0 & 0 \\ 0 & 0 & -1/r & 0 \end{bmatrix} \quad (5.1)$$

$$\mathbf{A}(\mathbf{x})_2 \equiv \begin{bmatrix} 0 & 1 & 0 & 0 \\ v^2/r^2 - \mu/r^3 & 0 & 0 & 0 \\ 0 & -v/r & 0 & 0 \\ 0 & 0 & -1/r & 0 \end{bmatrix} \quad (5.2)$$

$$\mathbf{A}(\mathbf{x})_3 \equiv \begin{bmatrix} 0 & 1 & 0 & 0 \\ -\mu/r^3 & 0 & v/r & 0 \\ 0 & 0 & -u/r & 0 \\ 0 & 0 & -1/r & 0 \end{bmatrix} \quad (5.3)$$

$$\mathbf{A}(\mathbf{x})_4 \equiv \begin{bmatrix} 0 & 1 & 0 & 0 \\ v^2/r^2 - \mu/r^3 & 0 & 0 & 0 \\ 0 & 0 & -u/r & 0 \\ 0 & 0 & -1/r & 0 \end{bmatrix} \quad (5.4)$$

These matrices are separately used in simulations of the fixed-final-state SDRE solution. The results of the simulations are shown in Table 5.

Table 5. Results of Experiment 1

Matrix	Run Time (min)	Max Thrust (N)	J (m^2/s^3)	Position Error (m)	Velocity Error (m/s)
$\mathbf{A}(\mathbf{x})_1$	1.443	56,317.39	2,752.53	0.00000320	0.00000934
$\mathbf{A}(\mathbf{x})_2$	1.267	54,755.46	2,755.50	0.00000301	0.00000879
$\mathbf{A}(\mathbf{x})_3$	0.970	56,230.35	2,757.50	0.00000304	0.00000885
$\mathbf{A}(\mathbf{x})_4$	1.076	54,550.60	2,760.95	0.00000286	0.00000834

For each $\mathbf{A}(\mathbf{x})$ matrix investigated, this table shows the simulation run time, the maximum thrust applied, the final cost J , and the final position and velocity errors. The simulation run times,

though difficult to measure accurately, show that different $\mathbf{A}(\mathbf{x})$ matrices can cause the calculations to run quicker or slower. This difference is more severe in some situations and is something that should be taken into consideration in the design of a controller. There is no significant difference in the final error in this experiment. However, there are differences in the values for maximum thrust and J . These differences should also be taken into consideration when designing a controller, depending on that which is more important.

From the results of this experiment, it is decided to use $\mathbf{A}(\mathbf{x})_1$ for the Phase 1 solution over the remainder of this research. This is because $\mathbf{A}(\mathbf{x})_1$ produces a lower J value, which means less fuel consumption.

5.1.2 Experiment 2

This experiment tests various initial conditions for Phase 1. The purpose of this is to see how different initial conditions affect the results and the behavior of the system. In the first part of the experiment, the values for initial range are varied to investigate how they affect the terminal states of Phase 1. The second part of the experiment investigates various values for initial altitude. The values tested in both parts are $\pm 1\%$, $\pm 2\%$, $\pm 3\%$, $\pm 4\%$, and $\pm 5\%$ of the nominal values. The results are shown in Table 6 and Table 7.

Table 6 shows the results for various values of initial range. All of the simulations reach the target values within an acceptable small amount of numerical error. The table shows that decreasing the initial range increases the values for maximum applied thrust and J , and increasing the initial range has the opposite effect. This means that starting a landing with a shorter range causes the control system to exert more energy to accomplish the landing.

However, the table also shows that an increased initial range causes slightly higher final error in position and velocity, and a decreased initial range has the opposite effect.

Table 6. Results for the first part of Experiment 2

% Difference in Initial Range	Max Thrust (N)	J (m^2/s^3)	Position Error (m)	Velocity Error (m/s)
-5%	61,849.79	2,840.51	0.00000266	0.00000775
-4%	60,741.85	2,820.94	0.00000276	0.00000804
-3%	59,634.59	2,802.36	0.00000286	0.00000835
-2%	58,528.06	2,784.76	0.00000297	0.00000867
-1%	57,422.31	2,768.15	0.00000308	0.00000899
0% (nominal)	56,317.39	2,752.53	0.00000320	0.00000933
+1%	55,213.36	2,737.90	0.00000332	0.00000969
+2%	54,110.30	2,724.26	0.00000345	0.00001006
+3%	53,008.26	2,711.61	0.00000358	0.00001043
+4%	51,907.34	2,699.95	0.00000371	0.00001082
+5%	50,807.62	2,689.28	0.00000385	0.00001123

In the second part of the experiment, as the initial altitude is varied, the velocity is adjusted to produce a circular orbit for each particular initial altitude. The purpose of this is to produce a more realistic situation. Therefore, the results for this part of the experiment are partially due to the changes in velocity.

Table 7 shows the results for the second part of this experiment. As before, all of the simulations reach the target values within an acceptable small amount of error, with the error increasing slightly with increased initial altitude. However, the initial altitude causes different effects on maximum applied thrust and J . Smaller values of initial altitude causes slightly larger values for maximum applied thrust and slightly smaller values for J , and larger values of initial altitude has the opposite effect. Therefore, starting the landing at a lower altitude reduces fuel consumption and final error, but it increases the required amount of thrust.

Table 7. Results for the second part of Experiment 2

% Difference in Initial Altitude	Max Thrust (N)	J (m^2/s^3)	Position Error (m)	Velocity Error (m/s)
-5%	56,336.27	2,751.05	0.00000317	0.00000924
-4%	56,332.43	2,751.34	0.00000318	0.00000927
-3%	56,328.62	2,751.64	0.00000318	0.00000928
-2%	56,324.84	2,751.93	0.00000319	0.00000930
-1%	56,321.10	2,752.23	0.00000319	0.00000932
0% (nominal)	56,317.39	2,752.53	0.00000320	0.00000933
+1%	56,313.70	2,752.83	0.00000320	0.00000935
+2%	56,310.06	2,753.13	0.00000321	0.00000937
+3%	56,306.44	2,753.44	0.00000322	0.00000939
+4%	56,302.85	2,753.74	0.00000322	0.00000940
+5%	56,299.30	2,754.05	0.00000323	0.00000942

Numerical experiments as shown in this section can be used to determine the optimal initial conditions for a landing. In addition, the results of this experiment add a level of confidence to the solution. The results show that, if the initial conditions are not exactly equal to the nominal conditions, the controller will still drive the lander to the desired final state, in the desired amount of time. However, of course, there will be limits based on the thrust capabilities of the lander and on the amount of fuel available.

5.1.3 Experiment 3

This experiment records the gain values from a simulation of Phase 1 using the fixed-final-state SDRE technique. It then uses these recorded gains in subsequent simulations to keep from having to calculate the gains in real time. The first purpose of this experiment is to test the robustness of the fixed-final-state SDRE technique. The second purpose is to demonstrate an alternate way of implementing the technique.

In this experiment, all of the cases of Experiment 2 are investigated again. However, this time all of the cases are investigated using recorded gain values of the nominal case. All of the $\mathbf{K}(t_c)$, $\mathbf{V}(t_c)$, and $\mathbf{P}(t_c)$ values are calculated along the trajectory for the nominal case and are recorded. These recorded values are then used for all of the other cases to find $\mathbf{u}(t)$ using Eq. (2.16)

The first part of this experiment investigates various values of initial range. The results are shown in Table 8. Even though the $\mathbf{K}(t_c)$, $\mathbf{V}(t_c)$, and $\mathbf{P}(t_c)$ values are only calculated for the nominal case, all of the simulations still reach the target values within an acceptable small amount of error. It is somewhat surprising that the final error values are so small in every case. Moreover, it is interesting to see that all values decrease with increasing initial range.

Table 8. Results for the first part of Experiment 3

% Difference in Initial Range	Max Thrust (N)	J (m^2/s^3)	Position Error (m)	Velocity Error (m/s)
-5%	61,849.79	2,842.40	0.00000420	0.00001224
-4%	60,741.85	2,822.47	0.00000404	0.00001180
-3%	59,634.59	2,803.52	0.00000387	0.00001128
-2%	58,528.06	2,785.55	0.00000367	0.00001070
-1%	57,422.31	2,768.55	0.00000345	0.00001005
0% (nominal)	56,317.39	2,752.53	0.00000320	0.00000933
+1%	55,213.36	2,737.49	0.00000294	0.00000856
+2%	54,110.30	2,723.43	0.00000265	0.00000775
+3%	53,008.26	2,710.36	0.00000236	0.00000690
+4%	51,907.34	2,698.27	0.00000207	0.00000605
+5%	50,807.62	2,687.17	0.00000180	0.00000525

The second part of this experiment investigates various values of initial altitude. The results are shown in Table 9. Again, all of the simulations reach the target values within an acceptable small amount of error. This, again, is surprising. The difference with this part of the experiment

is that, with increasing initial altitude, the maximum applied thrust and the J values increase, but the final error values decrease.

Table 9. Results for the second part of Experiment 3

% Difference in Initial Altitude	Max Thrust (N)	J (m^2/s^3)	Position Error (m)	Velocity Error (m/s)
-5%	56,287.46	2,751.15	0.00000332	0.00000968
-4%	56,293.40	2,751.43	0.00000330	0.00000962
-3%	56,299.35	2,751.70	0.00000327	0.00000955
-2%	56,305.34	2,751.98	0.00000325	0.00000947
-1%	56,311.35	2,752.25	0.00000322	0.00000940
0% (nominal)	56,317.39	2,752.53	0.00000320	0.00000933
+1%	56,323.45	2,752.81	0.00000318	0.00000926
+2%	56,329.54	2,753.09	0.00000315	0.00000919
+3%	56,335.65	2,753.37	0.00000313	0.00000912
+4%	56,341.79	2,753.66	0.00000310	0.00000904
+5%	56,347.95	2,753.94	0.00000308	0.00000898

The results of this experiment show that the technique is very robust, and that the calculated $\mathbf{K}(t)$, $\mathbf{V}(t)$, and $\mathbf{P}(t)$ matrices are good over a wide range of state values. This adds another level of confidence to the fixed-final-state SDRE technique. This experiment also shows that by storing the $\mathbf{K}(t)$, $\mathbf{V}(t)$, and $\mathbf{P}(t)$ matrices from a simulation, an alternate form of the technique can be implemented wherein the gains do not have to be calculated in real time.

5.1.4 Experiment 4

This experiment demonstrates a variation of the fixed-final-state SDRE technique. Velocity is removed from the fixed-end conditions, and the weighting matrix $\mathbf{S}(T)$ is formulated appropriately in order to minimize the magnitude of the final velocity. The values of $\mathbf{r}(T)$ and \mathbf{C} change to the following forms:

$$\mathbf{r}(T) \equiv \begin{bmatrix} r_f \\ \theta_f \end{bmatrix} = \begin{bmatrix} 1737130 \text{ m} \\ \pi/2 \text{ radians} \end{bmatrix}, \quad \mathbf{C} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 0 & 0 & 1 \end{bmatrix}$$

The $\mathbf{S}(T)$ matrix, from Eq. (2.10), changes to the form

$$\mathbf{S}(T) = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & (n_u) & 0 & 0 \\ 0 & 0 & (n_v) & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix}$$

in order to minimize the two velocity states. In this matrix, the n values are the “weights” that govern the amount of effort used to minimize the u and v velocities.

For this experiment, it is decided to define the matrix as

$$\mathbf{S}(T) = \begin{bmatrix} 0 & 0 & 0 & 0 \\ 0 & (n) & 0 & 0 \\ 0 & 0 & (n) & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \quad (5.5)$$

where the weighting values for both u and v are the same value n . Ten different values of n are used to form ten different $\mathbf{S}(T)$ matrices. Table 10 shows the values of n that are used, along with the names of the corresponding $\mathbf{S}(T)$ matrices. These matrices, along with the values of $\mathbf{r}(T)$ and \mathbf{C} above, are investigated in simulations to see how they affect the results of the fixed-final-state SDRE solution.

Table 10. Matrices investigated in Experiment 4

$\mathbf{S}(T)$:	$\mathbf{S}(T)_1$	$\mathbf{S}(T)_2$	$\mathbf{S}(T)_3$	$\mathbf{S}(T)_4$	$\mathbf{S}(T)_5$	$\mathbf{S}(T)_6$	$\mathbf{S}(T)_7$	$\mathbf{S}(T)_8$	$\mathbf{S}(T)_9$	$\mathbf{S}(T)_{10}$
n value:	1	10	10^2	10^3	10^4	10^5	10^6	10^7	10^8	10^9

The results of this experiment are shown in Table 11. The choice of $\mathbf{S}(T)$ affects all the values in the table. As the table shows, the position error is not affected significantly. However, when

n is small, such as in $\mathbf{S}(T)_1$, the velocity error is large. When n is large, such as in $\mathbf{S}(T)_8$, the velocity error is small. Increasing n more, as in $\mathbf{S}(T)_9$ and $\mathbf{S}(T)_{10}$, seems to increase the error slightly and decrease the maximum thrust by a small amount.

Table 11. Results of Experiment 4

$\mathbf{S}(T)$	Run Time (min)	Max Thrust (N)	J (m^2/s^3)	Position Error (m)	Velocity Error (m/s)
$\mathbf{S}(T)_1$	1.085	67,008.52	71,925.87	0.0002761	373.739135
$\mathbf{S}(T)_2$	1.105	60,538.74	139,704.65	0.0002034	165.743612
$\mathbf{S}(T)_3$	1.100	56,894.67	34,477.11	0.0001273	25.217778
$\mathbf{S}(T)_4$	1.121	56,370.86	6,281.12	0.0001046	2.659470
$\mathbf{S}(T)_5$	1.199	56,316.18	3,108.52	0.0000167	0.267123
$\mathbf{S}(T)_6$	1.272	56,310.54	2,786.64	0.0001508	0.026147
$\mathbf{S}(T)_7$	1.337	56,310.08	2,754.74	0.0001667	0.002105
$\mathbf{S}(T)_8$	1.253	56,310.12	2,753.47	0.0000883	0.000433
$\mathbf{S}(T)_9$	1.316	56,310.05	2,777.38	0.0001298	0.000705
$\mathbf{S}(T)_{10}$	1.391	56,309.93	3,079.61	0.0001596	0.000809

For this technique, it seems to take trial and error to decide on a value for $\mathbf{S}(T)$ that produces the desired results. From results of this experiment, it seems that $\mathbf{S}(T)_8$ is the best choice. It produces low error for both position and velocity, and it produces the lowest J value.

The results of this experiment show that using $\mathbf{S}(T)$ to minimize the magnitude of the final value of certain states is an alternative way to design the controller. Depending on the desired performance of the controller, this technique should be considered.

5.1.5 Comparison of the Techniques

This section compares the fixed-final-state SDRE solution, the ASRE solution, and the optimal variational solution for Phase 1 of the lunar landing. In the trajectory plots, a simulated

Apollo landing is also included. The purpose of these comparisons is to validate the fixed-final-state SDRE solution as a viable option for Phase 1.

Table 12 shows the results of the three solutions described previously for Phase 1. For each of the solutions, it lists the simulation run time, the maximum applied thrust, the cost value J , and the final error for position and velocity. All three solutions reach the target values within an acceptable small amount of error. The simulation run time for the SDRE solution is much less than an actual eleven-minute mission, so computational burden should not be a problem for a real mission. The maximum applied thrust values and the J values are very similar for the three solutions. The higher J values for the SDRE and ASRE solutions reflect the fact that these solutions are suboptimal. The approximating nature of these two techniques results in this higher cost, as would be expected.

Table 12. Results of the three solutions

Solution	Run Time (min)	Max Thrust (N)	J (m^2/s^3)	Position Error (m)	Velocity Error (m/s)
SDRE	1.443	56,317.39	2,752.53	0.00000320	0.00000934
ASRE	5.386	54,950.55	2,741.71	0.00000232	0.00000662
Optimal	0.118	54,372.93	2,734.53	0.00000000	0.00000000

Fig. 11 and Fig. 12 show the trajectories of the three solutions for Phase 1. Included in these plots is the trajectory for a simulated Apollo landing. The coordinate system for these plots is located at the center of the Moon.

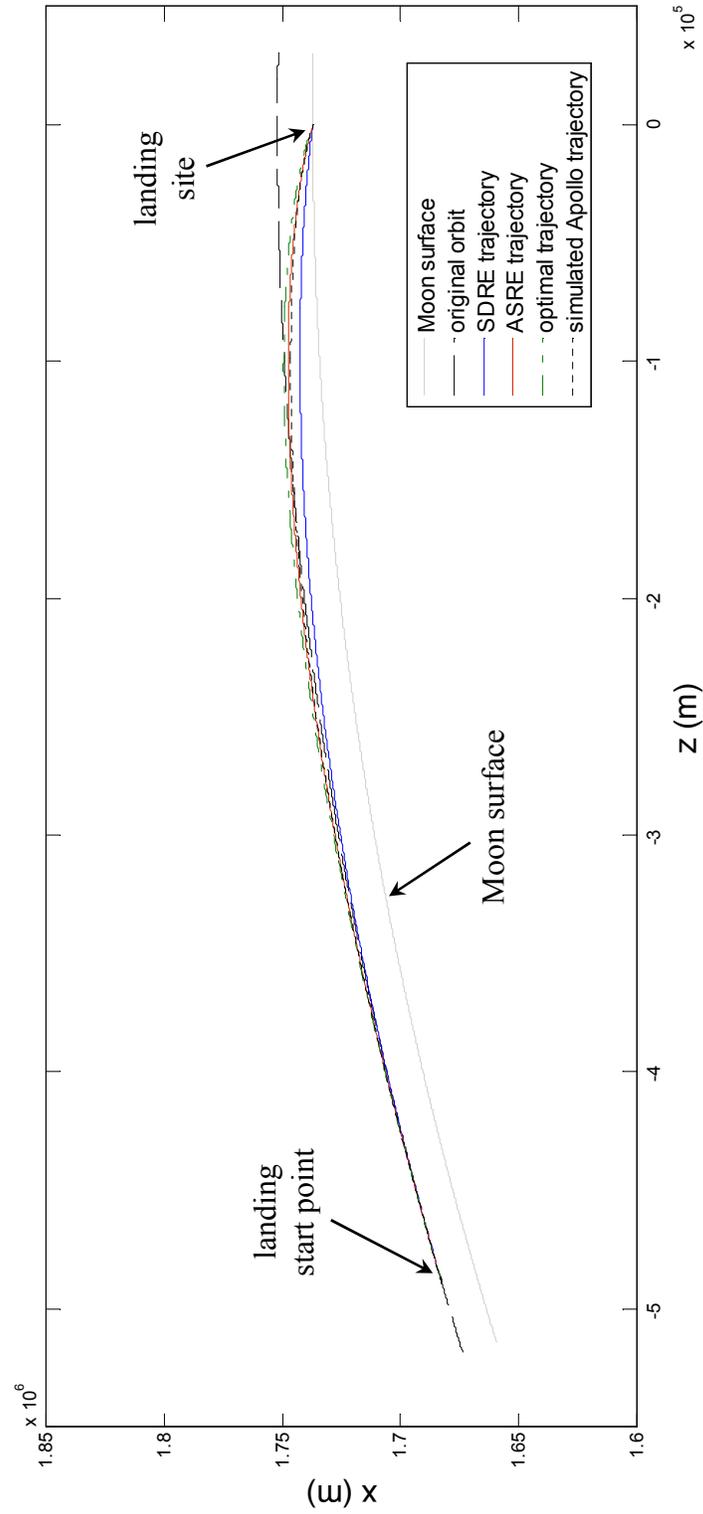


Fig. 11. Phase 1 landing trajectories of the various solutions

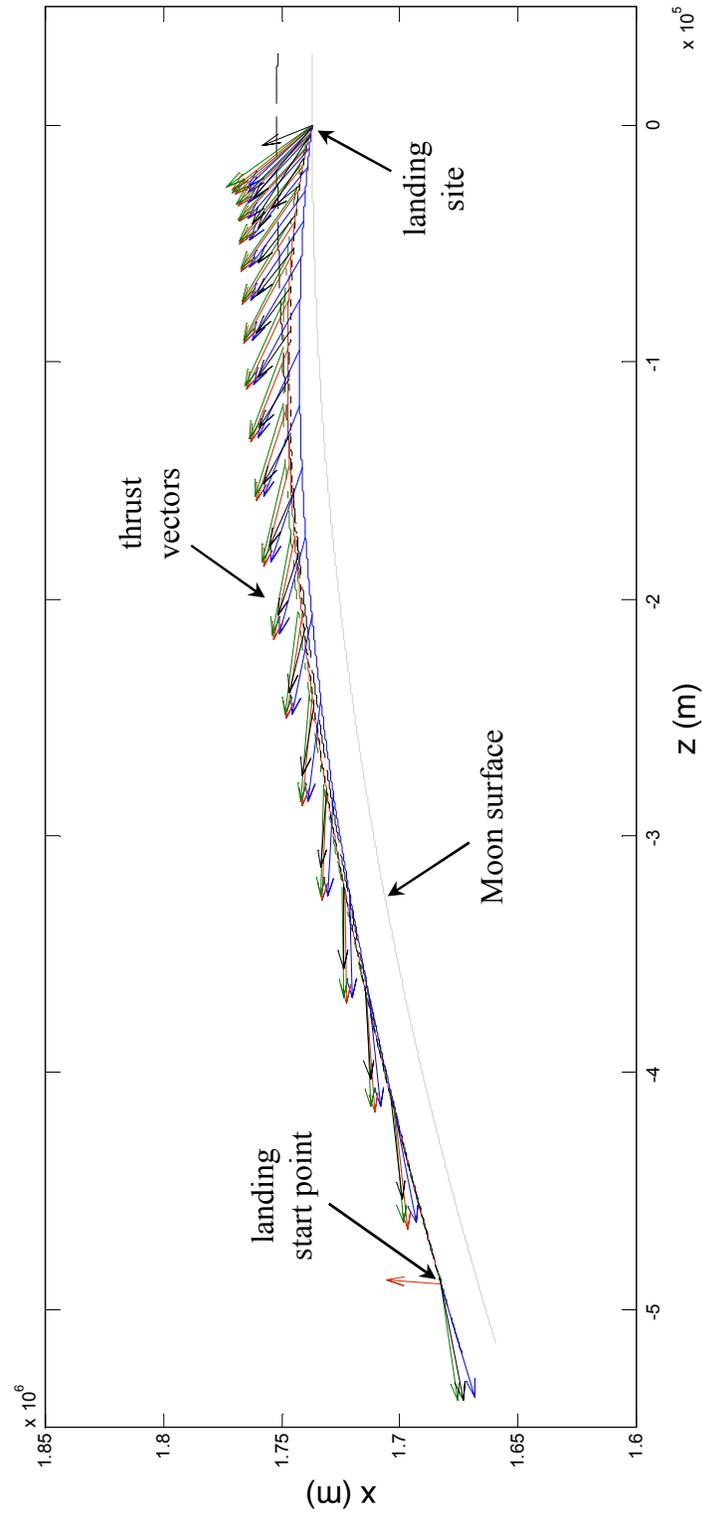


Fig. 12. Phase 1 landing trajectories of the various solutions, with thrust vectors

The trajectories in Fig. 11 show that all of the solutions, including that for Apollo, produce very similar results. The included thrust vectors in Fig. 12 show that the thrust profiles of all the solutions are very much alike.

With the coordinate system located at the center of the Moon, Fig. 13 and Fig. 14 show the terminal ends of the trajectories for Phase 1. These figures show how all of the solutions reach the intended target point. The ASRE trajectory is almost directly in line with the simulated Apollo trajectory. Again, the thrust vectors of Fig. 14 show the similarities of the thrust profiles.

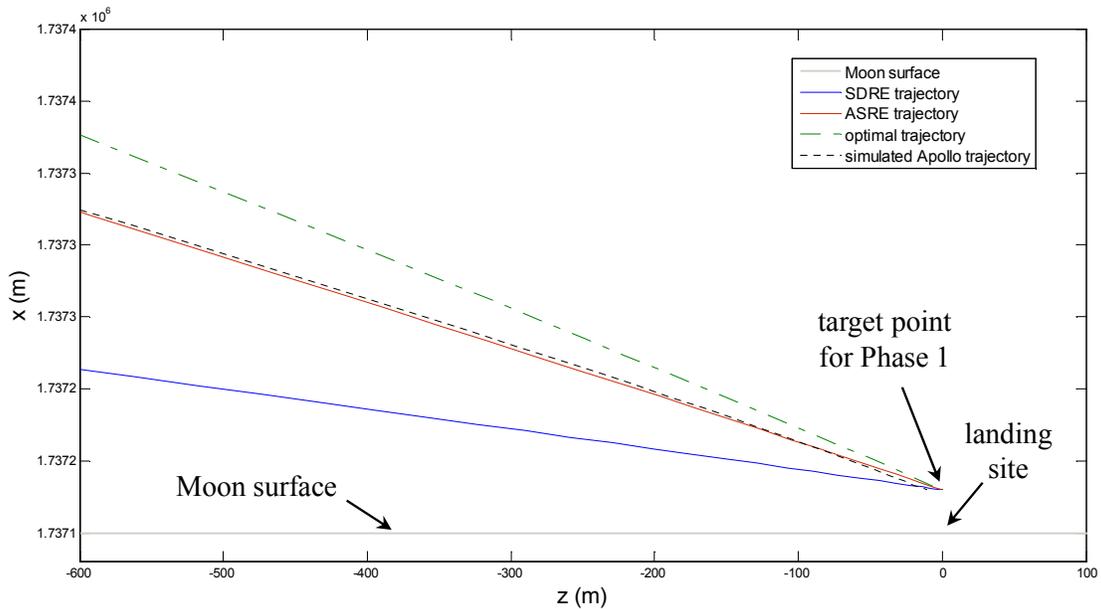


Fig. 13. Terminus of Phase 1 trajectories

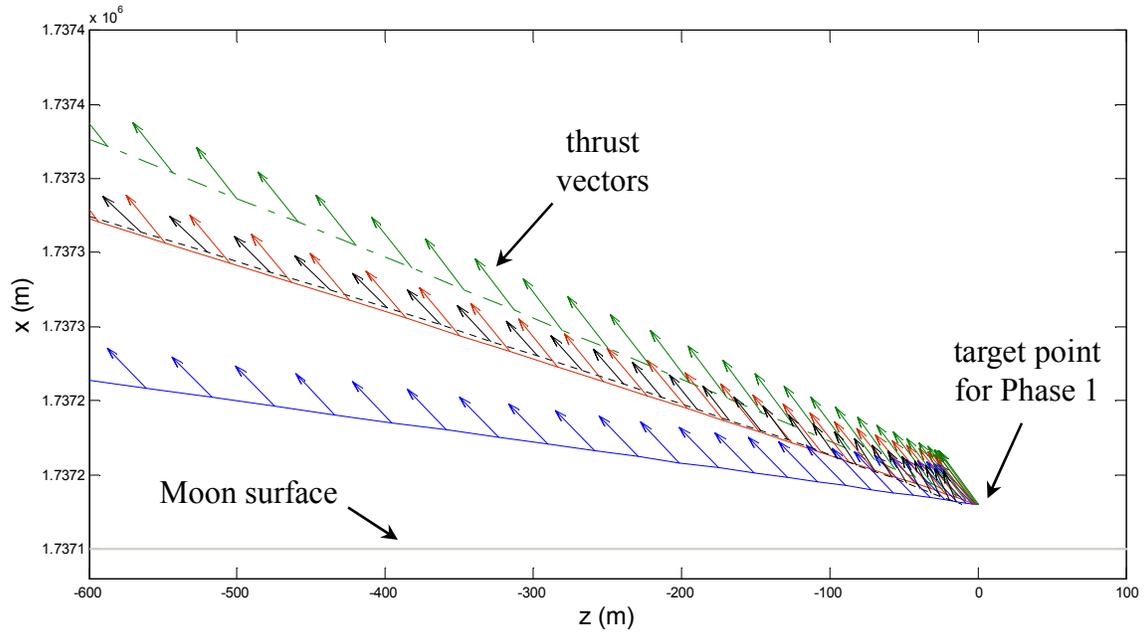


Fig. 14. Terminus of Phase 1 trajectories, with thrust vectors

Fig. 15 - Fig. 18 show plots of the states for the three optimal control solutions. The plots for r and u are very similar, but the plots for v and θ are identical.

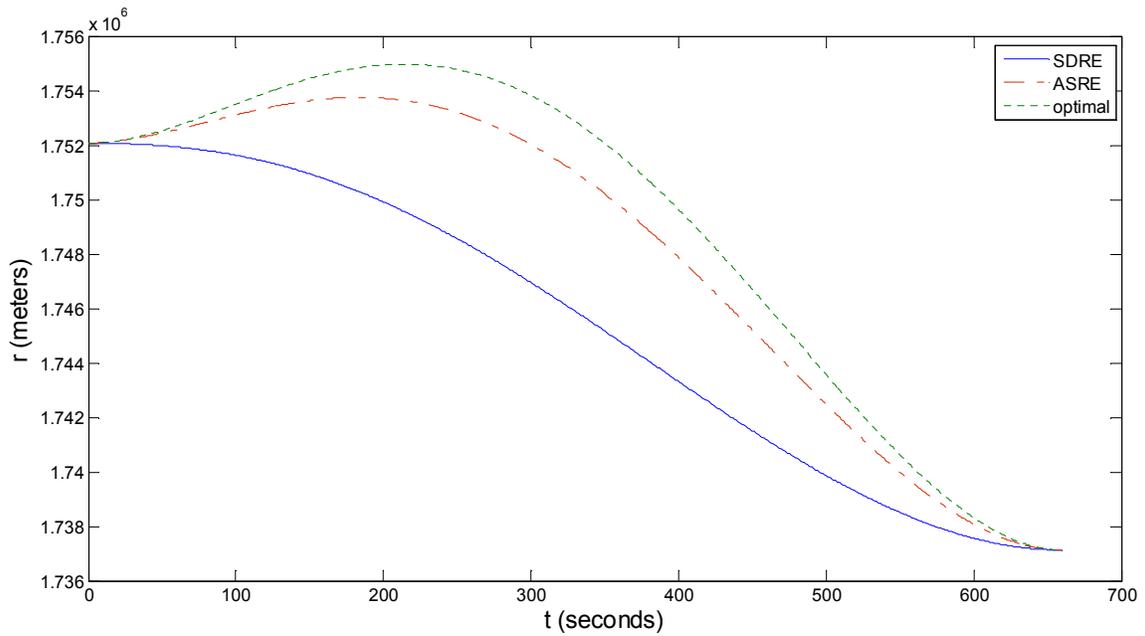


Fig. 15. Plot of r for the three solutions

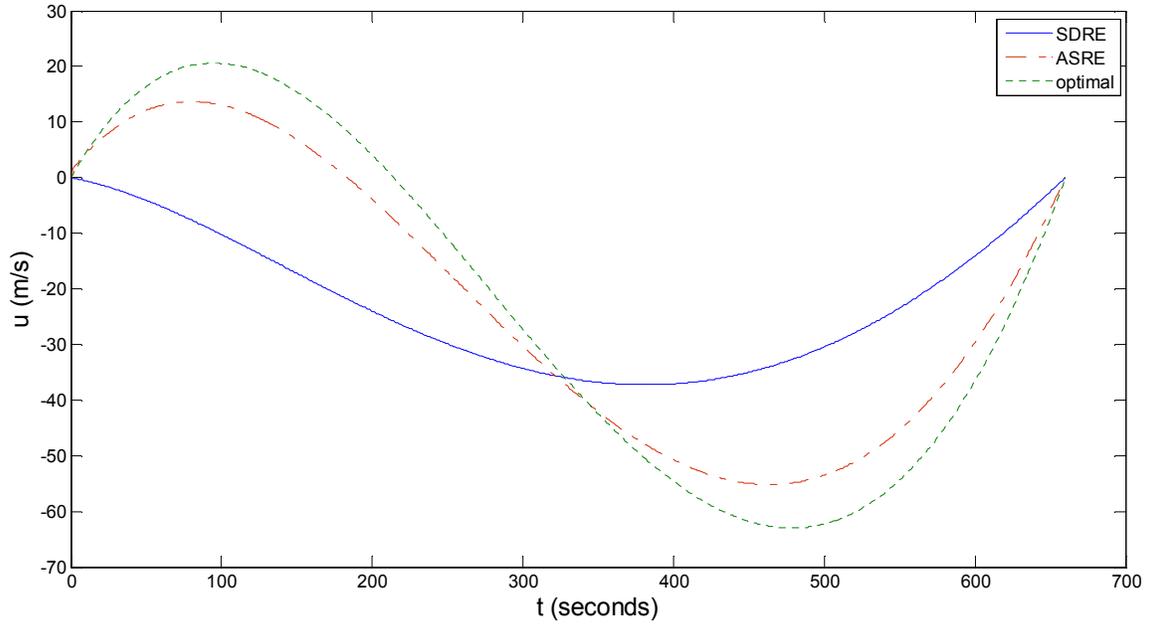


Fig. 16. Plot of u for the three solutions

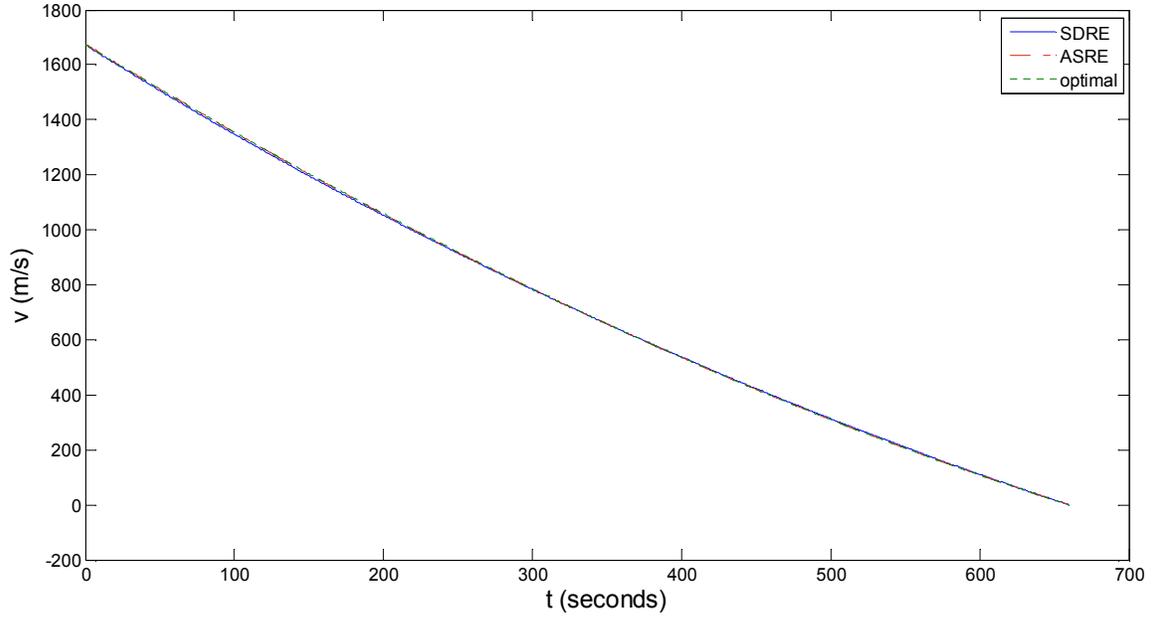


Fig. 17. Plot of v for the three solutions

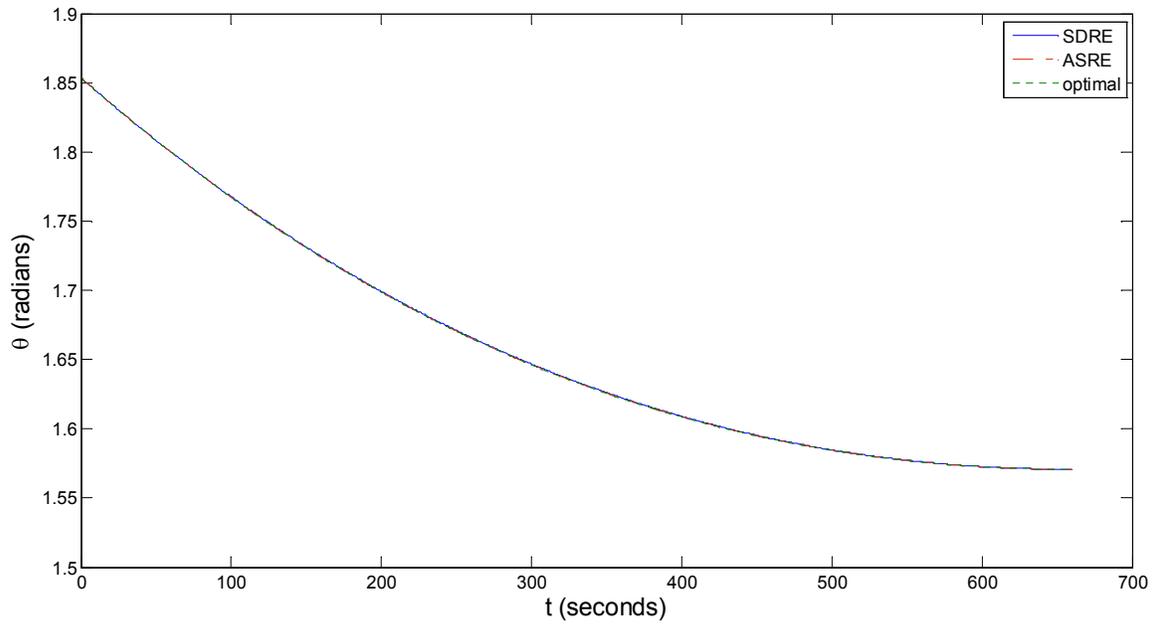


Fig. 18. Plot of θ for the three solutions

Fig. 19 - Fig. 22 show plots of the costates of the solutions. Section 2.2.1.3 gives a description of costate. Sections 4.2.2, 4.2.3, and 4.2.4 give explanations on how it is calculated. In these figures, the plots for the SDRE solution are closer than those are for the ASRE solution to matching the plots of the optimal solution. However, the differences between the costates of all three solutions are small.

Fig. 23 and Fig. 24 show plots of the input accelerations calculated for the three solutions. Fig. 23 shows plots for the input magnitude U , and Fig. 24 shows plots for the input angle ϕ . These plots are very similar for the three solutions. This is especially true for the plots of the input angle, which are almost identical.

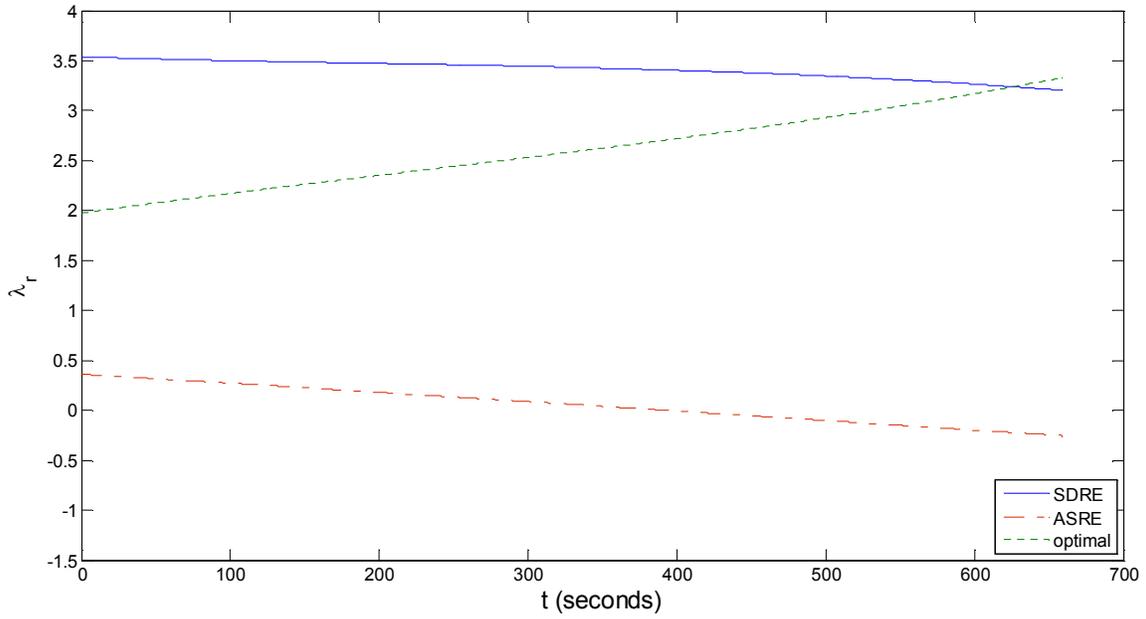


Fig. 19. Plot of λ_r for the three solutions

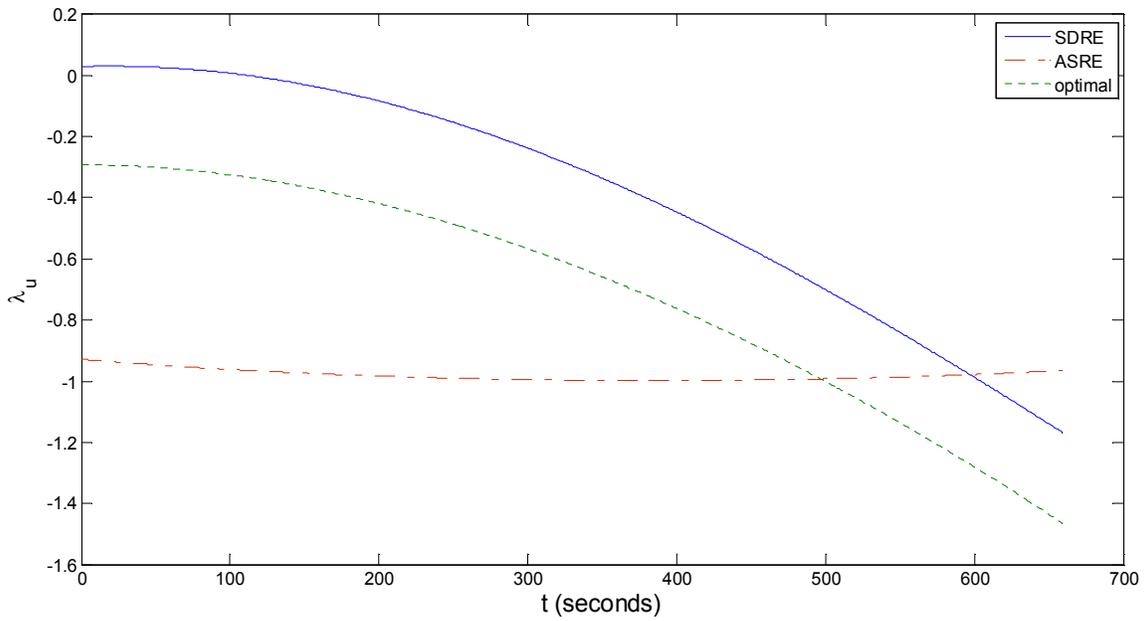


Fig. 20. Plot of λ_u for the three solutions

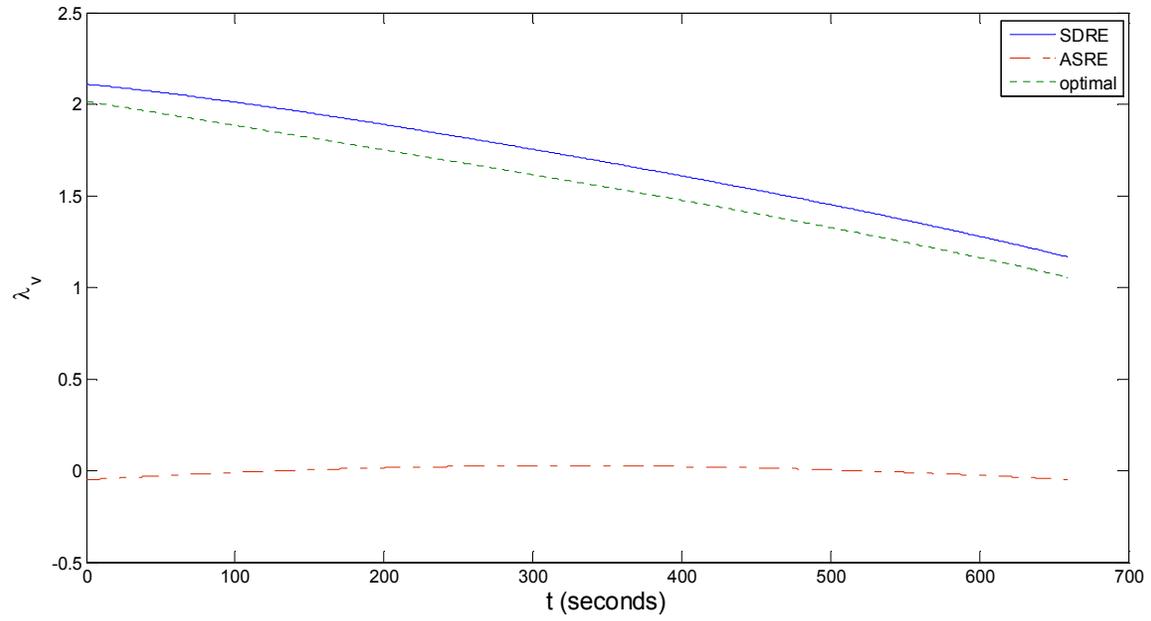


Fig. 21. Plot of λ_v for the three solutions

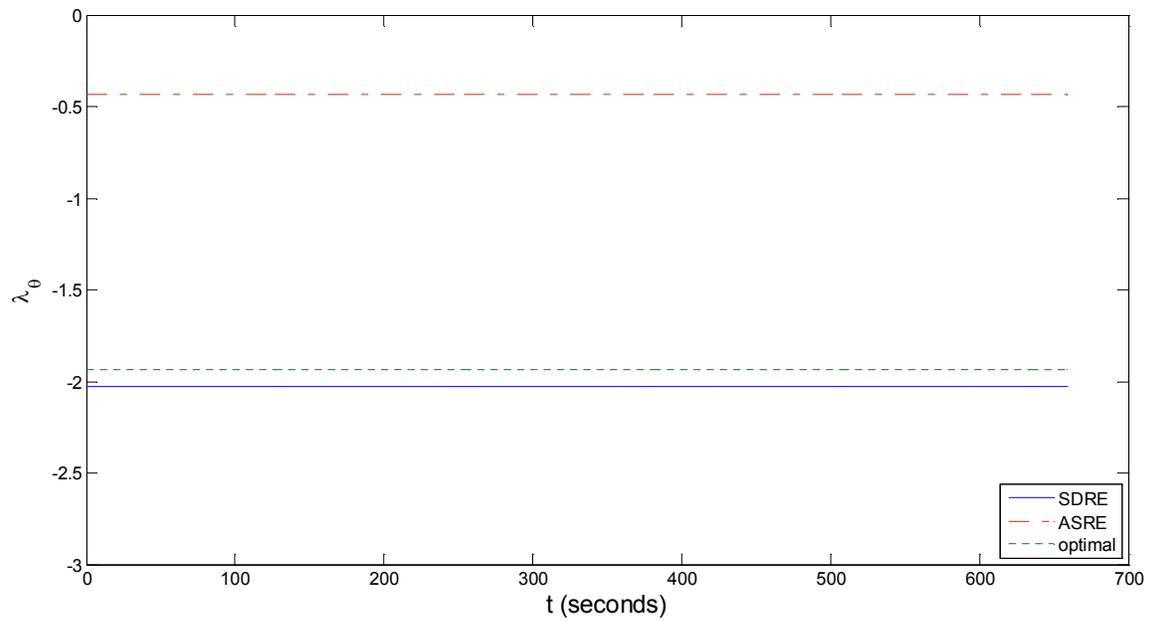


Fig. 22. Plot of λ_θ for the three solutions

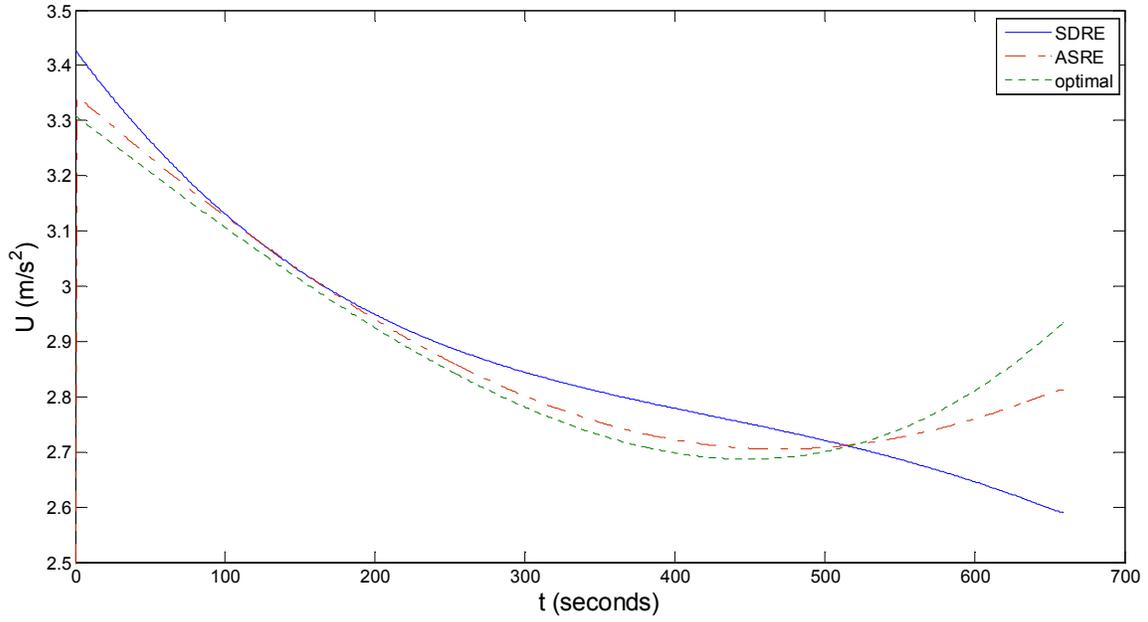


Fig. 23. Plot of U for the three solutions

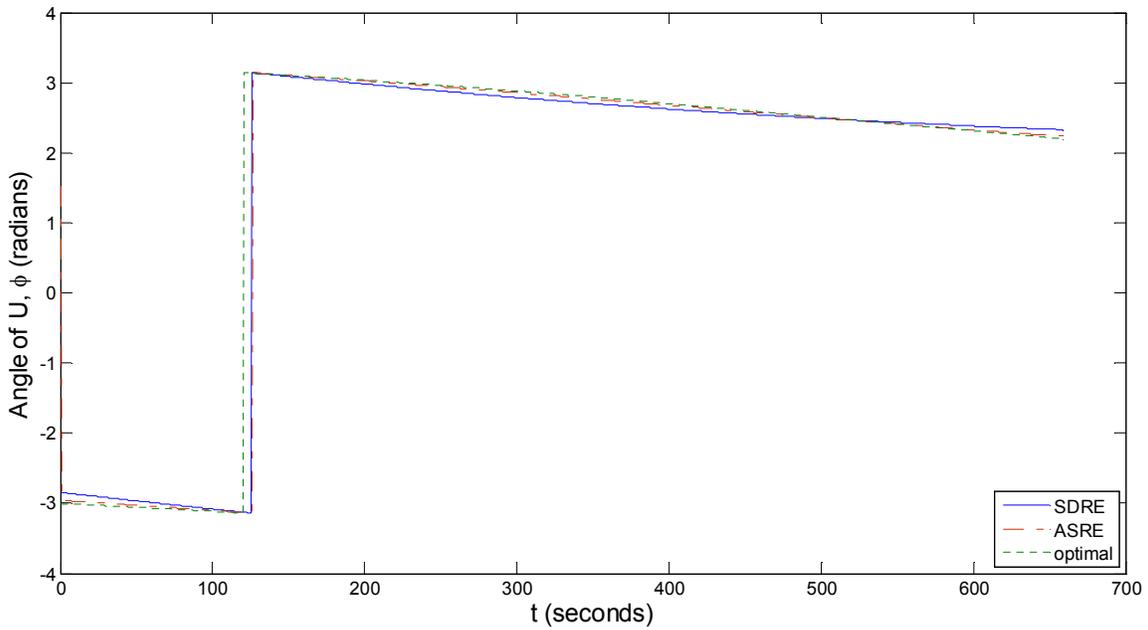


Fig. 24. Plot of the angle of U , ϕ , for the three solutions

Reversing the values of J with respect to time gives values of cost-to-go. Plots of cost-to-go are shown in Fig. 25. The plots for the three solutions are almost identical. As would be expected, the cost-to-go for all three solutions continually decreases over time.

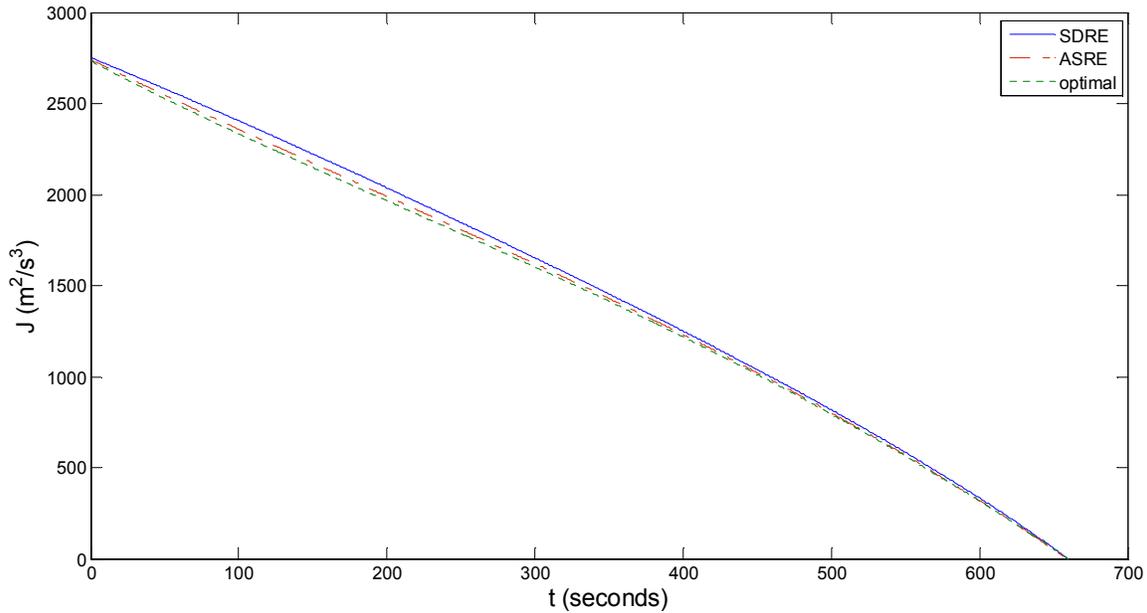


Fig. 25. Plot of cost-to-go, J , for the three solutions

5.2 Phase 2 Results

Phase 2, as described in Section 4.3, approximates the system as linear and then uses the linear fixed-final-state LQ control technique to finish the landing sequence. The final conditions of Phase 1 become the initial conditions for Phase 2. The controller then works to drive the lander to a soft landing on the lunar surface.

Table 13 lists the data from a simulation of this phase. For the thirty-second descent, the table shows position and velocity data in two-second intervals. It shows that the controller takes the lander from thirty meters above the landing site to a pinpoint soft landing at the surface.

Table 13. Data for Phase 2

Time (seconds)	Horizontal Position (meters)	Vertical Position (meters)	Horizontal Velocity (m/s)	Vertical Velocity (m/s)
0	0.000	30.000	0.000	0.000
2	0.052	29.615	-0.001	-0.376
4	0.051	28.532	-0.001	-0.698
6	0.048	26.859	-0.002	-0.966
8	0.044	24.704	-0.002	-1.180
10	0.040	22.176	-0.002	-1.340
12	0.035	19.382	-0.003	-1.445
14	0.029	16.430	-0.003	-1.497
16	0.024	13.429	-0.003	-1.495
18	0.019	10.486	-0.003	-1.439
20	0.014	7.707	-0.002	-1.330
22	0.009	5.201	-0.002	-1.168
24	0.006	3.073	-0.002	-0.952
26	0.003	1.428	-0.001	-0.684
28	0.001	0.371	-0.001	-0.365
30	0.000	0.000	0.000	0.000

Table 14 shows the final error for position and velocity. These values are satisfactory and show that the linear assumption for Phase 2 is valid.

Table 14. Results of Phase 2

Final Position Error (m)	Final Velocity Error (m/s)
0.00000000466	0.0000000528

CHAPTER 6

CONCLUSIONS

This research presents a new control method for landing on the Moon. The method divides the landing into two phases. The first phase uses a newly formulated technique for solving nonlinear problems, and the second phase uses a familiar technique for solving linear problems. The new nonlinear technique is the fixed-final-state SDRE method. Simulations of the two landing phases take a lander from lunar orbit to a gentle landing on the Moon's surface. The lander reaches the desired landing point in the desired amount of time, with pinpoint accuracy. This is accomplished without having to calculate the trajectory in advance.

Two other nonlinear optimal control techniques are also used to solve Phase 1. These are the ASRE technique and the indirect variational technique. The purpose of these additional solutions is to provide results that can be compared with those obtained from the fixed-final-state SDRE solution.

As the results show in Chapter 5, the new technique used for Phase 1 is accurate, reliable, and robust to the desired precision. Being a closed-loop feedback control technique, it has the ability to counteract unpredicted external inputs. It has a large degree of design flexibility because of the many choices for such things as the $\mathbf{A}(\mathbf{x})$ matrix, weighting matrices, time increment, and terminal constraints. In addition, the initial conditions and the target point for Phase 1 can be adjusted to create different trajectory profiles. In fact, the target point could be changed during a

landing, and the system would still be able to land accurately. For these reasons, this new technique should be considered for future missions to the Moon.

There are many possible experiments that could be performed in future research on the lunar landing method. One possibility is to have a craft to fly from one landing site to another. This may involve having a phase that would take the craft from the surface of the Moon to a determined altitude and then a second phase that would take the craft to the desired landing location. Another possibility for future investigation is variable sampling rate. Even though the simulations for this research use constant sampling rate, it should be easy to alter the method to vary the sample time increments over the course of the landing. If the system were to start the landing with long sample time increments and then make them shorter as the lander approaches the target point, this would require less computational burden and would result in a more efficient control routine.

REFERENCES

- [1] Bishop, R. H., and Azimov, D. M., 2008, "Enhanced Apollo Targeting for Lunar Landing," New Trends in Astrodynamics and Applications V, Anonymous The University of Texas at Austin.
- [2] Chomel, C. T., and Bishop, R. H., 2009, "Analytical Lunar Descent Guidance Algorithm," Journal of Guidance, Control, and Dynamics, 32(3) pp. 915-926.
- [3] Chomel, C. T., 2007, "Development of an Analytical Guidance Algorithm for Lunar Descent".
- [4] Christensen, D., and Geller, D., 2009, "Terrain-Relative and Beacon-Relative Navigation for Lunar Powered Descent and Landing," American Astronautical Society, AAS 09-057, Springfield, VA.
- [5] Guo, J., and Han, C., 2009, "Design of Guidance Laws for Lunar Pinpoint Soft Landing," American Astronautical Society, AAS 09-431, Springfield, VA.
- [6] Sostaric, R. R., 2007, "Powered Descent Trajectory Guidance and Some Considerations for Human Lunar Landing," 30th Annual AAS Guidance and Control Conference, Anonymous American Astronautical Society, San Diego, CA, AAS 07-051.
- [7] Klumpp, A.R., 1971, "Apollo Lunar-Descent Guidance," MIT Charles Stark Draper Laboratory, R-695, Cambridge, MA.
- [8] Bennett, F.V., 1972, "Apollo Experience Report- Mission Planning for Lunar Module Descent and Ascent," National Aeronautics and Space Administration, NASA TN D-6846, Washington, D.C.
- [9] Hoag, D.G., 1969, "Apollo Navigation, Guidance, and Control Systems," MIT Instrumentation Laboratory, E-2411, Cambridge, MA.
- [10] Nemeth, S., 2006, "Revisiting Apollo: Lunar Landing Guidance," AIAA-Houston Annual Technical Symposium, Anonymous United Space Alliance, LLC.
- [11] Curtis, H.D., 2005, "Orbital Mechanics for Engineering Students," Elsevier Ltd., Burlington, MA, pp. 673.
- [12] Brogan, W.L., 1991, "Modern Control Theory," Prentice Hall, Englewood Cliffs, N.J., pp. 653.
- [13] Bryson, A.E., and Ho, Y., 1975, "Applied Optimal Control : Optimization, Estimation, and Control," Hemisphere Pub. Corp., Washington, D.C., pp. 481.
- [14] Lewis, F.L., 1986, "Optimal Control," Wiley, New York, pp. 362.

- [15] Nelson, R.C., 1998, "Flight Stability and Automatic Control," McGraw-Hill, Boston, MA, pp. 441.
- [16] Palm, W.J., 1983, "Modeling, Analysis, and Control of Dynamic Systems," Wiley, New York, pp. 740.
- [17] Stevens, B.L., and Lewis, F.L., 2003, "Aircraft Control and Simulation," Wiley, Hoboken, NJ, pp. 664.
- [18] Holsapple, R., Venkataraman, R., and Doman, D.B., 2002, "A Modified Simple Shooting Method for Solving Two-Point Boundary-Value Problems," Air Force Research Laboratory, AFRL-VA-WP-TP-2002-327, Wright-Patterson Air Force Base, OH.
- [19] Trent, A., Venkataraman, R., and Doman, D. B., 2004, "Trajectory Generation Using a Modified Simple Shooting Method," 2004 IEEE Aerospace Conference, Anonymous IEEE, Big Sky, MT, 4, pp. 2723-2729.
- [20] Cloutier, J. R., 1997, "State-Dependent Riccati Equation Techniques: An Overview," American Control Conference, Anonymous pp. 932-936.
- [21] Cloutier, J. R., and Stansbery, D. T., 2002, "The Capabilities and Art of State-Dependent Riccati Equation-Based Design," American Control Conference, Anonymous pp. 86-91.
- [22] Mracek, C. P., and Cloutier, J. R., 2000, "Full Envelope Missile Longitudinal Autopilot Design using the State-Dependent Riccati Equation Method," Nonlinear Problems in Aviation and Aerospace, 11pp. 57-76.
- [23] Mracek, C. P., and Cloutier, J. R., 1998, "Control Designs for the Nonlinear Benchmark Problem Via the State-Dependent Riccati Equation Method," International Journal of Robust and Nonlinear Control, 8pp. 401-433.
- [24] Cimen, T., 2008, "State-Dependent Riccati Equation (SDRE) Control: A Survey," 17th World Congress of the International Federation of Automatic Control, Anonymous ROKETSAN Missiles Industries Inc, Ankara, Turkey.
- [25] Cimen, T., 2006, "Recent Advances in Nonlinear Optimal Feedback Control Design," 9th WSEAS International Conference on Applied Mathematics, Anonymous ROKETSAN Missiles Industries Inc, Ankara, Turkey.
- [26] Banks, H.T., Lewis, B.M., and Tran, H.T., 2003, "Nonlinear Feedback Controllers and Compensators: A State-Dependent Riccati Equation Approach," North Carolina State University, Center for Research in Scientific Computation, Raleigh, NC.
- [27] Beeler, S. C., Tran, H. T., and Banks, H. T., 2000, "Feedback Control Methodologies for Nonlinear Systems," Journal of Optimization Theory and Applications, 107(1) pp. 1-33.
- [28] Beeler, S.C., Tran, H.T., and Banks, H.T., 2000, "State Estimation and Tracking Control of Nonlinear Dynamical Systems," North Carolina State University, Center for Research in Scientific Computation, Raleigh, NC.
- [29] Bracci, A., Innocenti, M., and Pollini, L., 2006, "Estimation of the Region of Attraction for State-Dependent Riccati Equation Controllers," Journal of Guidance, Control, and Dynamics, 29(6) pp. 1427-1430.

- [30] Bradley, S.A., and Tsiotras, P., 2010, "A State-Dependent Riccati Equation Approach to Atmospheric Entry Guidance," American Institute of Aeronautics and Astronautics, AIAA 2010-8310.
- [31] Menon, P. K., Lam, T., Crawford, L. S., 2002, "Real-Time Computational Methods for SDRE Nonlinear Control of Missiles," American Control Conference, Anonymous Optimal Synthesis Inc, Los Altos, CA.
- [32] Shamma, J. S., and Cloutier, J. R., 2003, "Existence of SDRE Stabilizing Feedback," IEEE Transactions on Automatic Control, 48(3) pp. 513-517.
- [33] Yedavalli, R.K., Shankar, P., and Doman, D.B., 2003, "Combining State Dependent Riccati Equation Approach with Dynamic Inversion: Application to Control of Flight Vehicles," Air Force Research Laboratory, AFRL-VA-WP-TP-2003-300, Wright-Patterson Air Force Base, OH.
- [34] Zhang, Y., Agrawal, S. K., Hemanshu, P. R., 2005, "Optimal Control using State Dependent Riccati Equation (SDRE) for a Flexible Cable Transporter System with Arbitrarily Varying Lengths," 2005 IEEE Conference on Control Applications, Anonymous IEEE, Toronto, Canada, pp. 1063-1068.
- [35] Cimen, T., and Banks, S. P., 2004, "Global Optimal Feedback Control for General Nonlinear Systems with Nonquadratic Performance Criteria," Systems & Control Letters, 53(5) pp. 327-346.
- [36] Cimen, T., and Banks, S. P., 2004, "Nonlinear Optimal Tracking Control with Application to Super-Tankers for Autopilot Design," Automatica, 40pp. 1845-1863.
- [37] Betts, J.T., 2001, "Practical Methods for Optimal Control Using Nonlinear Programming," Society for Industrial and Applied Mathematics, Philadelphia, PA, pp. 190.
- [38] Ginsberg, J.H., 2008, "Engineering Dynamics," Cambridge University Press, Cambridge, New York, pp. 726.

APPENDIX A

DESCRIPTION OF CONTROL SYSTEMS AND OPTIMAL CONTROL

This section describes control systems [12, 15-17] and optimal control [13, 14]. The purpose of this material is to provide reference information.

A continuous-time dynamical system can be represented in a state space form [12] as

$$\dot{\mathbf{x}} = \mathbf{f}(\mathbf{x}, \mathbf{u}, t) \quad (\text{A.1})$$

where $\mathbf{x} \in \mathbb{R}^n$ is a vector of the system states, $\mathbf{u} \in \mathbb{R}^m$ is a vector of the inputs, and t is time.

This can be described in matrix form [12] to be

$$\dot{\mathbf{x}} = \mathbf{Ax} + \mathbf{Bu} \quad (\text{A.2})$$

where $\mathbf{A} \in \mathbb{R}^{n \times n}$ and $\mathbf{B} \in \mathbb{R}^{n \times m}$ are coefficient matrices. The diagram in Fig. 26 is a graphical representation of Equation (A.2) and a dynamical system.

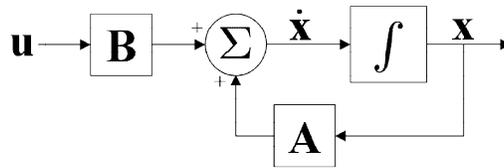


Fig. 26. Diagram of a system

A control system utilizes a control law to affect the input vector \mathbf{u} in such a way as to drive the state vector \mathbf{x} toward a desired set of values. In addition, the control law is designed to keep the system stable.

A control system can be either open-loop or closed-loop. An open-loop system contains a control law that is preprogrammed and generates \mathbf{u} values based only on time. This type of control system cannot compensate for unexpected disturbances. Fig. 27 shows a diagram of a general open-loop control system.

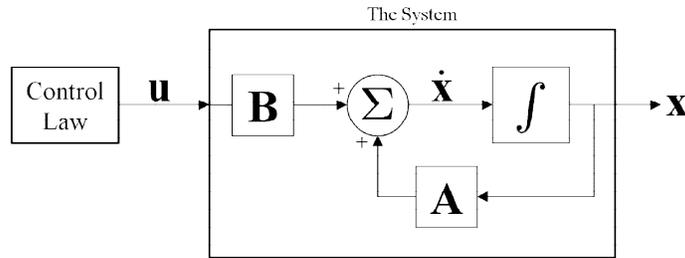


Fig. 27. Diagram of an open-loop control system

A closed-loop control system uses a control law that is based on time and state \mathbf{x} . This type of system, shown in Fig. 28, can compensate for unpredicted disturbances and other uncertainties.

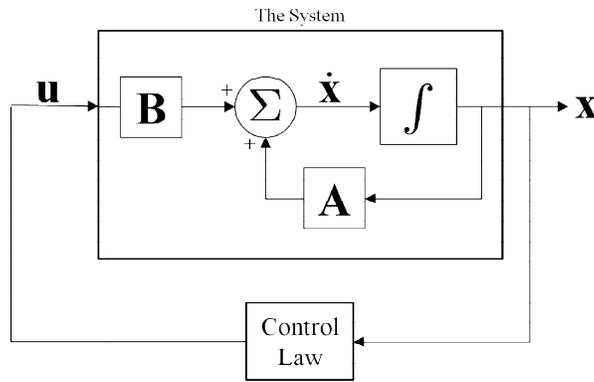


Fig. 28. Diagram of a closed-loop control system

Optimal control is a control law that seeks to achieve prescribed optimality goals. Typically, it works to minimize a cost functional J . For a finite-time system, $t \in [t_0, T]$, cost J can be defined by the quadratic functional [14] given as

$$J = \frac{1}{2} \mathbf{x}(T)^T \mathbf{S}(T) \mathbf{x}(T) + \frac{1}{2} \int_{t_0}^T (\mathbf{x}^T \mathbf{Q} \mathbf{x} + \mathbf{u}^T \mathbf{R} \mathbf{u}) dt \quad (\text{A.3})$$

with $\mathbf{S}(T) \geq 0$, $\mathbf{Q} \geq 0$, and $\mathbf{R} > 0$. The variables $\mathbf{S}(T)$, \mathbf{Q} , and \mathbf{R} are weighting matrices that can be chosen to achieve desired results. $\mathbf{S}(T)$ can be used to minimize the final state vector $\mathbf{x}(T)$, \mathbf{Q} can be used to minimize the states $\mathbf{x}(t)$ along the trajectory, and \mathbf{R} can be used to minimize the control input $\mathbf{u}(t)$ along the trajectory.

For an infinite-time system, $t \in [t_0, \infty)$, cost J can be defined by the quadratic functional [17] given as

$$J = \frac{1}{2} \int_0^{\infty} (\mathbf{x}^T \mathbf{Q} \mathbf{x} + \mathbf{u}^T \mathbf{R} \mathbf{u}) dt \quad (\text{A.4})$$

In this functional, \mathbf{Q} and \mathbf{R} are the same as defined above.

APPENDIX B

DERIVATION OF THE DYNAMICAL SYSTEM

This section shows the derivation of the applicable state equations used in the lunar landing problem of this research. For simplicity, the mass of the lander is assumed constant. Also, the landing problem is modeled in only two spatial dimensions. However, this is realistic because most of the action of a lunar landing occurs in a single plane. The unit vectors \mathbf{i}_r , \mathbf{i}_θ , and \mathbf{i}_z , for a cylindrical coordinate system, are used in the derivation.

To derive the state equations, first the position vector is defined as

$$\mathbf{r} = r \mathbf{i}_r \quad (\text{B.1})$$

Differentiating this with respect to time gives the expression for velocity to be

$$\mathbf{V} \equiv \dot{\mathbf{r}} = \frac{d}{dt}(r \mathbf{i}_r) = \frac{dr}{dt} \mathbf{i}_r + r \frac{d\mathbf{i}_r}{dt} \quad (\text{B.2})$$

Knowing that [38]

$$\frac{d\mathbf{i}_r}{dt} = \boldsymbol{\omega} \times \mathbf{i}_r$$

the following expression can be created:

$$r \frac{d\mathbf{i}_r}{dt} = \boldsymbol{\omega} \times r \mathbf{i}_r = \boldsymbol{\omega} \times \mathbf{r}$$

This changes the velocity expression in Equation (B.2) to

$$\mathbf{V} = \dot{r} \mathbf{i}_r + (\boldsymbol{\omega} \times \mathbf{r}) \quad (\text{B.3})$$

The radial velocity \mathbf{u} and the tangential velocity \mathbf{v} are now defined as

$$\mathbf{u} \equiv \dot{r} \mathbf{i}_r \quad \text{and} \quad \mathbf{v} \equiv \boldsymbol{\omega} \times \mathbf{r}$$

Therefore

$$u \mathbf{i}_r = \dot{r} \mathbf{i}_r \tag{B.4}$$

Differentiating the velocity expression in Equation (B.3) with respect to time gives an expression for acceleration as

$$\mathbf{a} \equiv \dot{\mathbf{V}} = \ddot{r} \mathbf{i}_r + \dot{r} (\boldsymbol{\omega} \times \mathbf{i}_r) + \dot{\boldsymbol{\omega}} \times \mathbf{r} + \boldsymbol{\omega} \times \dot{\mathbf{r}}$$

Expanding this gives

$$\mathbf{a} = \ddot{r} \mathbf{i}_r + 2(\boldsymbol{\omega} \times \dot{r} \mathbf{i}_r) + (\dot{\boldsymbol{\omega}} \times \mathbf{r}) + [\boldsymbol{\omega} \times (\boldsymbol{\omega} \times \mathbf{r})]$$

which can be represented as

$$\mathbf{a} = \dot{u} \mathbf{i}_r + 2(\boldsymbol{\omega} \mathbf{i}_z \times u \mathbf{i}_r) + (\dot{\boldsymbol{\omega}} \mathbf{i}_z \times r \mathbf{i}_r) + [\boldsymbol{\omega} \mathbf{i}_z \times (\boldsymbol{\omega} \mathbf{i}_z \times r \mathbf{i}_r)] \tag{B.5}$$

With the tangential velocity vector \mathbf{v} shown to be

$$\mathbf{v} = v \mathbf{i}_\theta = \boldsymbol{\omega} \mathbf{i}_z \times r \mathbf{i}_r = \omega r \mathbf{i}_\theta$$

and angular velocity defined as $-\dot{\theta}$, the following expression can be formed:

$$\omega = -\dot{\theta} = v / r \tag{B.6}$$

Now, Equation (B.5) can be simplified in terms of u and v to give

$$\mathbf{a} = \dot{u} \mathbf{i}_r + \frac{2uv}{r} \mathbf{i}_\theta + \left(\dot{v} - \frac{uv}{r} \right) \mathbf{i}_\theta + -\frac{v^2}{r} \mathbf{i}_r$$

This shows that radial acceleration can be given as

$$a_r = |a \mathbf{i}_r| = \dot{u} - \frac{v^2}{r} \tag{B.7}$$

and tangential acceleration can be given as

$$a_t = |a \mathbf{i}_\theta| = \dot{v} + \frac{uv}{r} \quad (\text{B.8})$$

For an orbital body, the total radial acceleration depends on the radial distance r , the gravitational parameter μ , and the radial input acceleration U_r by the expression [11]

$$a_r = -\frac{\mu}{r^2} + U_r$$

Combining this with Equation (B.7) gives

$$\dot{u} - \frac{v^2}{r} = -\frac{\mu}{r^2} + U_r \quad (\text{B.9})$$

Tangential acceleration can be equated to be the tangential input acceleration as

$$a_t = U_t$$

Combining this with Equation (B.8) gives

$$\dot{v} + \frac{uv}{r} = U_t \quad (\text{B.10})$$

From Eqs. (B.4), (B.6), (B.9), and (B.10), the nonlinear set of state equations for the dynamical system are obtained to be

$$\dot{r} = u \quad (\text{B.11})$$

$$\dot{u} = \frac{v^2}{r} - \frac{\mu}{r^2} + U_r \quad (\text{B.12})$$

$$\dot{v} = -\frac{uv}{r} + U_t \quad (\text{B.13})$$

$$\dot{\theta} = -\frac{v}{r} \quad (\text{B.14})$$